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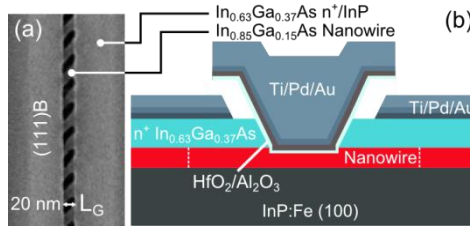
# High-Frequency InGaAs Tri-gate MOSFETs with $f_{\max}$ of 400 GHz

C. B. Zota, F. Lindelöw, L.-E. Wernersson and E. Lind

We report on extremely scaled down tri-gate RF MOSFETs utilizing lateral nanowires as the channel, with gate length and nanowire width both of 20 nm. These devices exhibit simultaneous extrapolated  $f_i$  and  $f_{\max}$  of 275 and 400 GHz at  $V_{DS} = 0.5$  V, which is the largest combined  $f_i$  and  $f_{\max}$ , as well as the largest  $f_{\max}$  reported for all III-V MOSFET.

**Introduction:** Tri-gate (or non-planar) MOSFETs for RF-applications are motivated by that the use of a high-k oxide, rather than a semiconductor barrier (as in HEMTs) allows for higher gate capacitance in the MOSFET [1-2]. Furthermore, the tri-gate architecture improves short-channel effects, allowing for shorter gate length,  $L_G$ , without degradation of performance due to short-channel effects. Both these points enable higher ideal transconductance,  $g_m$ , in MOSFETs compared to HEMTs, assuming similar electron mobility. In fact, state-of-the-art III-V MOSFET devices exhibit  $g_m$  larger than that of record HEMTs, although they presently do not allow RF-compatible device designs [3-5].

In this work, we present RF-compatible tri-gate  $\text{In}_{0.85}\text{Ga}_{0.15}\text{As}$  MOSFETs utilizing lateral nanowires (NWs) as the channel. Compared to our previous work, we have here further scaled down device dimensions,  $L_G$  and nanowire width,  $W_{NW}$  [6]. This enables higher  $g_m$  at  $V_{DS} = 0.5$  V, which significantly improves  $f_i/f_{\max}$  from 220/305 GHz to 275/400 GHz. The combined  $f_i$  and  $f_{\max}$ , as well as the  $f_{\max}$  of these devices represent the highest reported values for all III-V MOSFETs.



**Fig. 1** Device fabrication and device materials and design

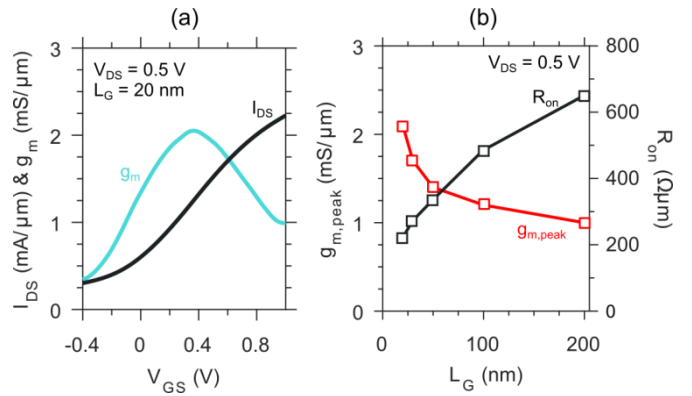
a SEM image of the device after contact regrowth,  $L_G$  is defined as the distance between  $n^+$  contacts. (111)B denotes the crystal facet of the contact layer.

b Schematic figure of the fabricated device.

**Fabrication:** The device fabrication process is similar to what has been described elsewhere [7]. The device channel consists of 200 lateral  $\text{In}_{0.85}\text{Ga}_{0.15}\text{As}$  nanowires, formed by selective area MOCVD growth on (100)  $\text{InP:Fe}$  (S.I.) substrate, split over two gate fingers. The nanowire width is 20 nm, and the height is 11 nm. The S/D highly doped regions are formed by a second MOCVD growth step of 40 nm  $n^+$   $\text{In}_{0.63}\text{Ga}_{0.37}\text{As}/100$  nm  $\text{InP}$  with in-situ Sn-doping ( $N_D = 5 \times 10^{19} \text{ cm}^{-3}$ ) in the doped layer (Fig. 1a). Subsequently, 1 nm/5 nm  $\text{Al}_2\text{O}_3/\text{HfO}_2$  is deposited by ALD and  $\text{Ti/Pd/Au}$  by thermal evaporation, forming the gate stack. The regrown 100 nm  $\text{InP}$  is selectively etched by an HCl solution leaving a T-gate. S/D and pad metallization of  $\text{Ti/Pd/Au}$  completes the process (Fig. 1b).

**Results:** Fig. 2a shows transfer characteristics of a device with  $L_G = 20$  nm measured at DC with a Keithley 4200 semiconductor characterization system. All data is normalized to the total gated periphery of the NWs (7  $\mu\text{m}$ ). At  $V_{DS} = 0.5$  V, peak  $g_m$  is 2.1  $\text{mS}/\mu\text{m}$ . Fig. 2b shows the scaling behaviour of peak  $g_m$  and on-resistance  $R_{on}$  versus  $L_G$ .  $R_{on}$  reaches 220  $\Omega\mu\text{m}$  at  $L_G = 20$  nm. The total access resistance is estimated to 130  $\Omega\mu\text{m}$  from transmission line measurements.

RF-measurements were performed at 40 MHz to 67 GHz with an Agilent E8361A vector network analyser. On-chip pad de-embedding as well as off-chip two-port load-reflect-reflect-match calibration was performed. The total pad capacitances were approximately 20 fF.

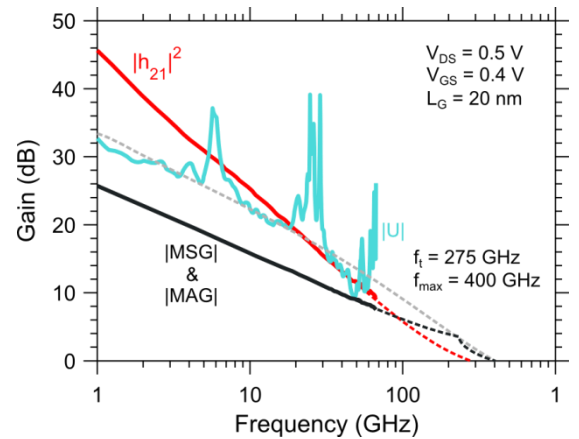


**Fig. 2** Device characteristics at DC.

a Transfer characteristics of a  $L_G = 20$  nm device.

b Scaling behaviour of peak  $g_m$  and  $R_{on}$  versus  $L_G$ .

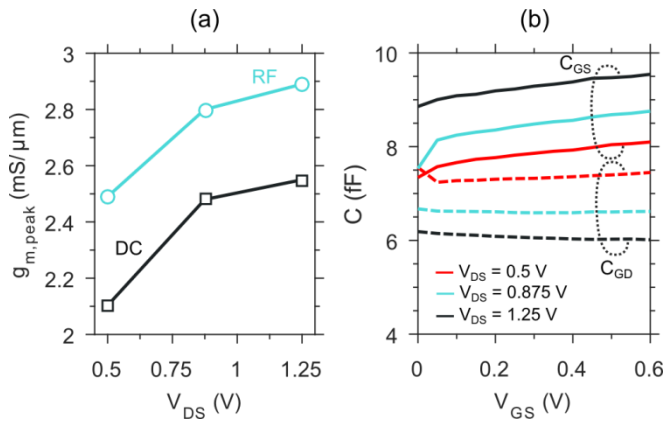
A small-signal model was determined from the measured S-parameters, with a good fit to the measurement data [8]. Fig. 3 shows measured and modelled (dashed traces) unilateral power gain  $|U|$ , current gain  $|h_{21}|^2$  and maximum available/stable gain ( $|MAG|$  and  $|MSG|$ ) for a device with  $L_G = 20$  nm. Extrapolated cut-off frequency  $f_i$  is 275 GHz and maximum oscillation frequency  $f_{\max}$  is 400 GHz.



**Fig. 3** Measured and modelled (dashed traces) gain of an  $L_G = 20$  nm device at  $V_{DS} = 0.5$  V.

The small-signal model, which is similar to that in [6], includes both the effect of border traps in the oxide, and impact ionization. Border traps are modelled using the distributed border trap model in [9]. Border traps introduce a frequency-dependency to  $g_m$  and  $g_d$ , as well as a frequency-dependent oxide loss, and explain the -10 dB slope of  $|U|$  versus  $f$  [10]. Fig. 4a shows  $g_{m,\text{peak}}$  for an  $L_G = 20$  nm device extracted from the small-signal model at DC and 67 GHz (RF).  $g_{m,\text{peak}}$  increases by approximately 13% in the latter case, to a maximum of 2.9  $\text{mS}/\mu\text{m}$  at  $V_{DS} = 1.25$  V, which is attributable to that trap responses are partially disabled at high frequency.

The effective gate resistance is  $\sim 5 \Omega$ , and the source and drain resistances are  $\sim 2 \Omega$ . The gate-to-source and gate-to-drain capacitances,  $C_{GS}$  and  $C_{GD}$ , are shown in Fig. 4b. At  $V_{DS} = 0.5$  V, the total gate capacitance  $C_{GS} + C_{GD}$  is 15 fF at peak  $g_m$ . This includes both the parasitic capacitance from the source and drain gate overlaps, and the intrinsic gate capacitance. The latter is estimated as  $C_{\text{gg,int}} = (2/3)WLC_{\text{ox}}/(C_q + C_{\text{ox}})$ , with the quantum capacitance  $C_q = q^2 m^*/\pi\hbar^2$ , which is  $\sim 2$  fF with  $m^* = 0.04m_0$ . Thus, RF-performance is primarily limited by the parasitic overlap capacitance, which can be lowered by implementation of source and drain spacers.

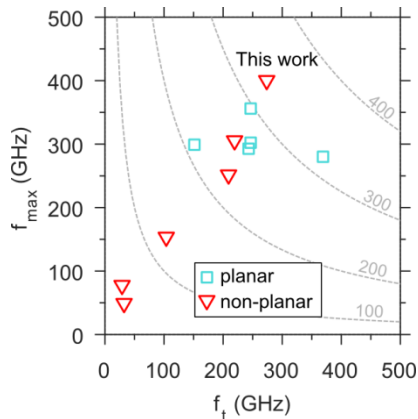


**Fig. 4** Peak  $g_m$  and capacitances

a Peak  $g_m$  measured at both 40 MHz (DC) and 67 GHz (RF), for an  $L_G = 20$  nm device.

b Gate-to-source,  $C_{GS}$ , and gate-to-drain,  $C_{GD}$ , capacitances measured at different  $V_{DS}$ .

Fig. 5 shows a benchmark of  $f_t$ ,  $f_{max}$  and the geometric mean  $\sqrt{f_t \times f_{max}}$  (dashed traces) for state-of-the-art III-V MOSFETs [11-18]. The geometric mean is 330 GHz for these devices, which is the highest reported value for a III-V MOSFET. Squares show planar devices, and triangles show non-planar devices.



**Fig. 5** Benchmark of RF-performance for III-V MOSFETs

Squares show planar devices, triangles show non-planar devices.  $V_{DS}$  and  $L_G$  varies between devices, but is 0.5 V and 20 nm, respectively, for this work. Dashed traces show the geometric mean.

**Conclusion:** We have demonstrated  $L_G = 20$  nm  $\text{In}_{0.85}\text{Ga}_{0.15}\text{As}$  tri-gate MOSFETs with record high-frequency performance,  $f_t = 275$  GHz and  $f_{max} = 400$  GHz at  $V_{DS} = 0.5$  V.

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