

#### A State Space Model of a Multimachine Power System

Lindahl, Sture

1971

Document Version: Publisher's PDF, also known as Version of record

Link to publication

Citation for published version (APA): Lindahl, S. (1971). A State Space Model of a Multimachine Power System. (Research Reports TFRT-3037). Department of Automatic Control, Lund Institute of Technology (LTH).

Total number of authors:

General rights

Unless other specific re-use rights are stated the following general rights apply:

Copyright and moral rights for the publications made accessible in the public portal are retained by the authors and/or other copyright owners and it is a condition of accessing publications that users recognise and abide by the legal requirements associated with these rights.

• Users may download and print one copy of any publication from the public portal for the purpose of private study

- You may not further distribute the material or use it for any profit-making activity or commercial gain
   You may freely distribute the URL identifying the publication in the public portal

Read more about Creative commons licenses: https://creativecommons.org/licenses/

Take down policy

If you believe that this document breaches copyright please contact us providing details, and we will remove access to the work immediately and investigate your claim.

# A STATE SPACE MODEL OF A MULTIMACHINE POWER SYSTEM.

STURE LINDAHL

REPORT 7118 NOVEMBER 1971
DIVISION OF AUTOMATIC CONTROL
LUND INSTITUTE OF TECHNOLOGY

#### A STATE SPACE MODEL OF A MULTIMACHINE POWER SYSTEM.

#### S. Lindahl

#### ABSTRACT.

The equations for a multimachine power system are derived. The model includes hydro turbines as well as steam turbines and boilers. The nonlinear equations are derived from basic physical laws. They are linearized to obtain a linear state space model, which is valid for small perturbations about an operating point. A method of obtaining the equations on standard state space form is proposed. Except matrix multiplication the method only requires the inversion of one n×n matrix (complex) and n 3×3 matrices (real), where n is the number of generating plants.

This work has been supported by the Swedish Board for Technical Development under Contract 71-50/U33.

TA	BLE OF CONTENTS	Page
1.	INTRODUCTION	1
2.	DESCRIPTION OF THE POWER SYSTEM	4
3.	SYNCHRONOUS MACHINE AND EXCITER	8
	3.1. The Ideal Synchronous Machine	8
,	3.2. Transformation from a 3-Phase Machine to a 2-Phase Machine	11
	3.3. Transformation from a 2-Phase Machine to a dq_Machine	16
	3.4. The Air-Gap Torque	22
	3.5. Linearized Equations for the Synchronous Generator	24
	3.6. Basic Equations for the Exciter	28
4.	TRANSMISSION NETWORK	29
	4.1. Selection of Angular References	29
	4.2. Transformation of Network Equations	30
5.	PRIME MOVERS	35
	5.1. Hydro Trubines	36
	5.2. Boilers and Steam Turbines	39
6.	CONSTRUCTION OF SYSTEM MATRICES	42
	6.1. Differential Equations for Rotor Angles (x <sub>1</sub> )	<b>4</b> 4
	6.2. Differential Equations for Rotor Angular Velocities (x.)	<b>4</b> 5

### Table of Contents (contd.)

	6.3.	Differential Equations for	
		Field Flux Linkage (x3)	49
	6.4.	Differential Equations for	
		Armature d-Axis Flux Linkage	
		(x <sub>4</sub> )	50
	6.5.	Differential Equations for	
		Armature q-Axis Flux Linkage	
		(x <sub>5</sub> )	52
	6.6.	Differential Equations for	
		Excitation Volatage (x <sub>6</sub> )	52
	6.7.	Differential Equations for	
		Prime Mover State Variables	
		(x <sub>7</sub> )	53
	6.8.	Structure of the Complete	
		System	54
7.	REFER	RENCES	56

APPENDIX: List of Symbols

#### 1. INTRODUCTION.

A power system consists of several plants, a large distribution network and a variety of consumers. To analyze the performance of the system, power engineers have developed computer programs for simulation of multimachine power systems [1], [3], [8], [9]. Methods of analyzing the stability of the linearized model, describing small perturbations about an operating point, has also been developed [2], [4], [5]. Methods of improving the performance of a power system have also been proposed, but these methods often assume that:

- o a single generator is connected to an infinite bus,
- o the mechanical input to the generator is constant.

Under the above assumptions the voltage regulator becomes a single-input single-output system, and the classical methods can be applied to design the voltage regulator. If we remove one of the above assumptions, the model becomes multivariable and classical control theory does not provide a systematic method of designing the regulators. Of course, simulation can be used to find suitable tuning of the regulators. One drawback of such simulations is the amount of computing time required.

Modern control theory enables us to handle multivariable systems, described by a set of first order linear differential equations

 $\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{u} \tag{1.1}$ 

where x is the state vector, u the control vector and

A, B the coefficient matrices. To apply linear-quadratic control theory we assume that the performance can be described by

$$V = \int_{0}^{\infty} \left[ x^{T}(s) Q_{1}x(s) + u^{T}(s) Q_{2}u(s) \right] ds$$
 (1.2)

where  $Q_1$  is a symmetric nonnegative definite matrix, and  $Q_2$  is a symmetric positive definite matrix. The problem is to find a control u(t), such that the loss function V is minimized. The solution to the problem is given by the linear time-invariant feed-back.

$$u(t) = -Lx(t) \tag{1.3}$$

where

$$L = Q_2^{-1} B^{T} S (1.4)$$

The matrix S is the symmetric nonnegative definite solution of the stationary Riccati equation

$$A^{T}S + SA + Q_{1} - SBQ_{2}^{-1}B^{T}S = 0$$
 (1.5)

The control signals are linear combinations of all state variables. To implement such a controller it is necessary to transmit the whole state vector to every plant and this may not be realistic. Since we obtain a solution with all possible feedbacks we have a yard-stick to evaluate the importance of feeding certain variables from one station to another. We also have the tools to analyze various suboptimal strategies. In any case it is necessary to simulate the nonlinear equations, describing the system, and using

the actual control law. If the chosen control law works it is immaterial that we have found it by applying linear-quadratic control theory. This approach is feasible only if the total amount of computing time is less than the computing time required for straightforward simulation.

In this report we derive the equations for a multimachine power system with hydro turbines as well as steam turbines and boilers as prime movers. We also propose a method of building up the system matrices A and B in (1.1).

In Section 2 we describe the power system configuration. The basic equations for the synchronous machine are derived in Section 3. The synchronous machines are connected to the transmission network and in Section 4 we consider the transmission network. In Section 5 we present the nonlinear equations for the prime movers. Finally we describe the method of obtaining the system matrices.

#### 2. DESCRIPTION OF THE POWER SYSTEM.

The basic system studied is shown in Fig. 2.1.

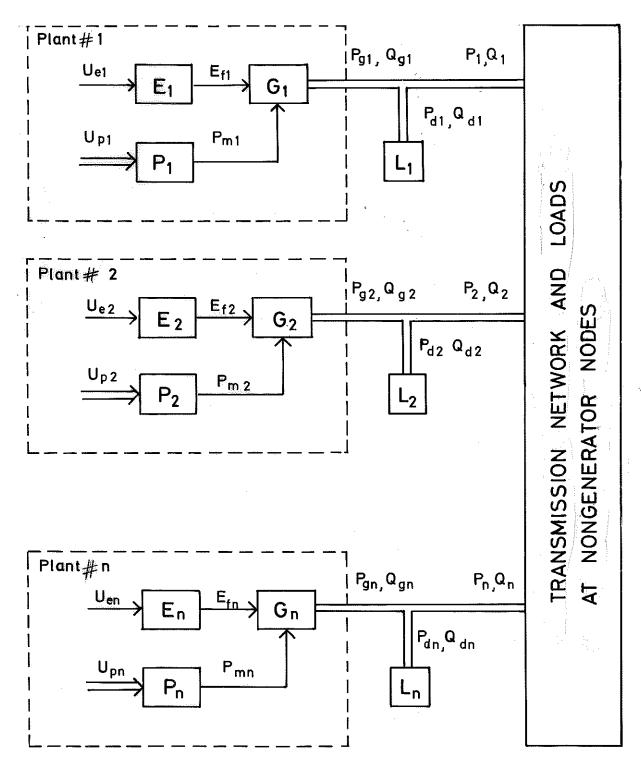


Fig. 2.1 - Schematic diagram of a power system consisting of n generating plants. Each plant is composed of a prime mover (P<sub>i</sub>), a synchronous machine (G<sub>i</sub>) and an excitation system (E<sub>i</sub>).

The system consists of a linear multiport lumped parameter electrical transmission network, n generating plants and local loads at the generator nodes. The loads at nongenerator nodes are already included in the transmission network.

At every generator node active and reactive power, denoted  $P_i$  and  $Q_i$  respectively, is fed into the transmission network. The active and reactive power demand at the generator nodes is denoted by  $P_{di}$  and  $Q_{di}$  in Fig. 2.1. The sum of injected power  $(P_i,Q_i)$  and local demand  $(P_{di},Q_{di})$  is equal to the generated power  $(P_{gi},Q_{gi})$ .

The generating plants consist of a synchronous machine  $(G_i)$ , an excitation system  $(E_i)$  and a prime mover  $(P_i)$ . The inputs to the synchronous machine are the field voltage  $(E_{fi})$  and the mechanical power  $(P_{mi})$ . The input signal to the excitation system is denoted by  $U_{ei}$  in Fig. 2.1.

The input signal (s) to the prime mover is denoted by  $U_{\mbox{\footnotesize{pn}}}$ 

The network is treated as if it was in steady-state operating condition. The alternating node voltages and current are represented by the complex quantities  $\hat{V}$  and  $\hat{I}$  respectively. The transmission network is assumed to be completely described by the complex nodal admittance matrix Y. The nodal admittance equation can be written

$$\tilde{I} = \tilde{Y} \cdot \tilde{V} \tag{2.1}$$

where all nonsynchronous loads are represented by constant admittances and incorporated into Ŷ by eliminating all nongenerator nodes.

The n complex equations (2.1) are separated into 2n real equations as proposed in [10].

$$\begin{bmatrix} \mathbf{i}_{D1} \\ \mathbf{i}_{D2} \\ \vdots \\ \mathbf{i}_{Dn} \\ \mathbf{i}_{Q2} \\ \vdots \\ \mathbf{i}_{Qn} \end{bmatrix} = \begin{bmatrix} g_{11} & g_{12} & \cdots & g_{1n} & -b_{11} & -b_{12} & \cdots & -b_{1n} \\ g_{21} & g_{22} & \cdots & g_{2n} & -b_{21} & -b_{22} & \cdots & -b_{2n} \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ g_{n1} & g_{n2} & \cdots & g_{nn} & -b_{n1} & -b_{n2} & \cdots & -b_{nn} \\ b_{11} & b_{12} & \cdots & b_{1n} & g_{11} & g_{12} & \cdots & g_{1n} \\ b_{21} & b_{22} & \cdots & b_{2n} & g_{21} & g_{22} & \cdots & g_{2n} \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ b_{n1} & b_{n2} & \cdots & b_{nn} & g_{n1} & g_{n2} & \cdots & g_{nn} \end{bmatrix} \begin{bmatrix} \mathbf{v}_{D1} \\ \mathbf{v}_{D2} \\ \vdots \\ \mathbf{v}_{Qn} \end{bmatrix}$$

$$(2.2)$$

This equation is written symbolically as

$$\begin{bmatrix} \mathbf{I}_{\mathbf{D}} \\ \mathbf{I}_{\mathbf{Q}} \end{bmatrix} = \begin{bmatrix} \mathbf{G}_{\mathbf{N}} & -\mathbf{B}_{\mathbf{N}} \\ \mathbf{B}_{\mathbf{N}} & \mathbf{G}_{\mathbf{N}} \end{bmatrix} \begin{bmatrix} \mathbf{V}_{\mathbf{D}} \\ \mathbf{V}_{\mathbf{Q}} \end{bmatrix}$$
 (2.3)

The synchronous machines are described by the set of Park's equations [6], [7] given in Section 3. The excitation systems are modelled by first order dynamics.

Each synchronous machine is connected to either a hydro turbine or a boiler and steam turbine. The equations for the prime movers are derived in Section 5.

The following variables are used as state variables:

- Rotor angle
- o Rotor angular velocity

- Flux linkage of field winding
- Flux linkage of d-axis winding
- Flux linkage of q-axis winding
- Excitation voltage
- Water speed (hydro plants)
- Steam pressure (steam plants)

In comparison with the other variables the flux linkages of d- and q-axis winding ( $\psi_d$  and  $\psi_q$  respectively) changes very rapidly and the differential equations for  $\psi_d$  and  $\psi_q$  are often approximated by algebraic equations. In this case  $\psi_d$  and  $\psi_q$  are not contained in the state vector but can be expressed as a linear combination of the state variables.

The following variables are used as input variables:

- Excitation input
- Gate opening (hydro plants)
- Steam valve setting (steam plants)
- Fuel flow (steam plants)

#### 3. SYNCHRONOUS MACHINE AND EXCITER.

In this section we rederive Park's equations [6], [7] for the synchronous machine. Often these equations are rederived under the assumption that the machine is in steady-state, but used for the machine in transient state. Our task is to find a set of equations valid for transient as well as steady-state conditions, and this is one reason to rederive Park's equations. The material in this section is mainly based on [12].

The first step in this process is to transform the original 3-phase machine to a 2-phase machine with the same magnetomotive force (mmf). Then we transform the 2-phase machine to the dq-machine applying a second linear transformation which removes the time-varying inductances of the 2-phase machine.

#### 3.1. The Ideal Synchronous Machine.

The windings of a 3-phase 2-pole synchronous machine are shown in Fig. 3.1. On the stator there are the three distributed a-c windings r, s and t, one in each phase. They are symbolized by the correspondingly labeled concentrated coils. The magnetic axes of the phase windings coincide with the coil axes.

The d-c field winding, f, is on the rotor. The effect of the damper windings is included in a general damping term as described in Section 3.5.

The rotor has two axes of symmetry, the polar, or direct axis d and the interpolar, or quadrature, axis q. The magnetic flux paths have different permeances in the two directions of axis. The d and q

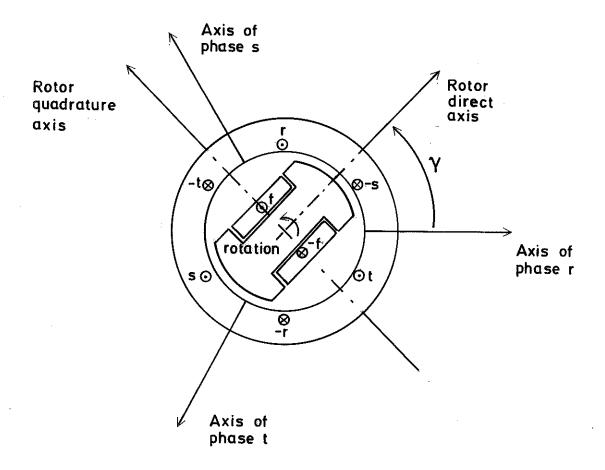


Fig. 3.1 - An idealized synchronous machine.

axes revolve with the rotor, while the magnetic axes of the three stator phases remain fixed.

In deriving the basic equations required for modelling of a synchronous machine it is assumed that:

- A 3.1) The stator windings are sinusoidally distributed around the air-gap as far as the mutual effects between them and the rotor are concerned.
- A 3.2) The stator winding self\_and mutual\_inductances vary sinusoidally as the rotor revolves,

and are of the form  $a+b \cdot \cos 2\gamma$  and  $c+b \cdot \cos (2\gamma-2\pi/3)$  respectively, where a,b,c and d are constants.

#### A 3.3) Saturation and hysteresis are negligible.

The circuits r, s, t and f have their own resistance and their own self-inductance and mutual inductance with respect to every other circuit. The script letter  $\ell$  with appropriate subscripts is used to denote these inductances for any value of  $\gamma$ . In terms of the self and mutual inductances  $\ell$ , the flux linkages are

$$\begin{bmatrix} \psi_{r} \\ \psi_{s} \\ \psi_{t} \\ \end{bmatrix} = \begin{bmatrix} \ell_{rr} & \ell_{rs} & \ell_{rt} & \ell_{rf} \\ \ell_{sr} & \ell_{ss} & \ell_{st} & \ell_{sf} \\ \ell_{tr} & \ell_{ts} & \ell_{tt} & \ell_{tf} \\ \ell_{fr} & \ell_{fs} & \ell_{ft} & \ell_{ff} \end{bmatrix} \begin{bmatrix} i_{r} \\ i_{s} \\ i_{t} \\ i_{f} \end{bmatrix}$$
(3.1)

or symbolically

$$\Psi_{rstf} = L_{rstf}I_{rstf}$$
 (3.2)

In (3.1) all inductances except  $l_{ff}$  are functions of  $\gamma$  and thus time-varying. We observe that assumption (A 3.3) is necessary for (3.1).

The following expressions for induced emf are valid for the 3-phase machine:

$$v_r = r_a i_r + d\psi_r / dt$$
 (3.3)

$$v_{s} = r_{a}i_{s} + d\psi_{s}/dt \qquad (3.4)$$

$$v_{t} = r_{a}i_{t} + d\psi_{t}/dt$$
 (3.5)

$$v_f = r_f i_f + d\psi_f / dt$$
 (3.6)

or symbolically

$$V_{rstf} = R_{rstf}I_{rstf} + p\Psi_{rstf}$$
 (3.7)

where

px = dx/dt

r<sub>a</sub> - armature resistance

 $r_{f}$  - field resistance

## 3.2. Transformation from a 3-phase Machine to a 2-phase Machine.

In this section we transform the 3-phase machine to a 2-phase machine with the same mmf distribution. We do not change the geometry of the iron circuits but permit the number of effective turns to change.

For the transformation from 3-phase to 2-phase we require that:

- R 3.1) The instantaneous value of the mmfs must be equal.
- R 3.2) The currents, voltages and flux linkages must be transformed with the same matrix.
- R 3.3) The instantaneous power ( $I^{T}V$ ) must be invariant.

Denote the effective number of turns/phase with  $\rm N_3$  and  $\rm N_2$  and divide the mmf into components on the

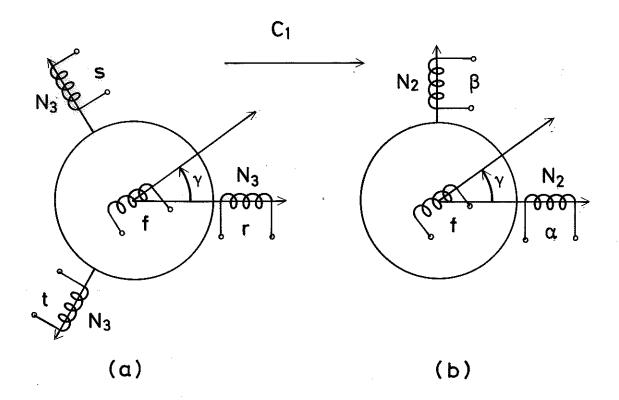


Fig. 3.2 - Transformation from 3-phase to 2-phase.

The original 3-phase machine (a). The resulting 2-phase machine with the same mmf distribution (b).

 $\alpha$ - and  $\beta$ -axes. R 3.1) now gives:

$$N_2 i_\alpha = N_3 (i_r - i_s / 2 - i_t / 2)$$
 (3.8)

$$N_2 i_\beta = N_3 (\sqrt{3} i_s/2 - \sqrt{3} i_t/2)$$
 (3.9)

$$N_{oi_{o}} = N_{3}(i_{r}+i_{s}+i_{t})$$
 (3.10)

$$i_f = i_f \tag{3.11}$$

The current io does not produce any field in the airgap and is associated with the stator leakage inductance.

Under balanced 3-phase conditions  $i_0$  is zero. Equations (3.8) to (3.11) can be written in matrix form:

$$\begin{bmatrix} \mathbf{i}_{\alpha} \\ \mathbf{i}_{\beta} \\ \mathbf{i}_{o} \\ \mathbf{i}_{f} \end{bmatrix} = \begin{bmatrix} K & -K/2 & -K/2 & 0 \\ 0 & \sqrt{3}K/2 & -\sqrt{3}K/2 & 0 \\ K_{1} & K_{1} & K_{1} & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} \mathbf{i}_{r} \\ \mathbf{i}_{s} \\ \mathbf{i}_{t} \\ \mathbf{i}_{f} \end{bmatrix}$$
(3.12)

where  $K = N_3/N_2$  and  $K_1 = N_3/N_0$ .

Equation (3.12) can be written in symbolical form as:

$$I_{\alpha\beta\circ f} = C_1 I_{rstf} \tag{3.13}$$

Requirement R 3.2) now gives:

$$V_{\alpha\beta\text{of}} = C_1 V_{\text{rstf}} \tag{3.14}$$

Requirement R 3.3) and equations (3.13) and (3.14) further yield:

$$P_{\alpha\beta\text{of}} = I_{\alpha\beta\text{of}}^{T} V_{\alpha\beta\text{of}} = I_{\text{rstf}}^{T} C_{1}^{T} C_{1}^{V} V_{\text{rstf}} =$$

$$= P_{\text{rstf}} = I_{\text{rstf}}^{T} V_{\text{rstf}}$$

Hence

$$C_1^T C_1 = I$$
 (3.15)

Condition (3.15) implies

$$K = \sqrt{2/3}$$

$$K_1 = 1/\sqrt{3}$$

Summing up we find that the linear transformation  $\mathbf{C}_1$  is given by

$$I_{\alpha\beta\circ f} = C_1 I_{rstf}$$
 (3.16)

$$V_{\alpha\beta\circ f} = C_1 V_{rstf}$$
 (3.17)

$$\Psi_{\alpha\beta\circ f} = C_1 \Psi_{rstf} \tag{3.18}$$

$$C_{1} = \sqrt{\frac{2}{3}} \begin{bmatrix} 1 & -1/2 & -1/2 & 0 \\ 0 & \sqrt{3}/2 & -\sqrt{3}/2 & 0 \\ 1/\sqrt{2} & 1/\sqrt{2} & 1/\sqrt{2} & 0 \\ 0 & 0 & 0 & \sqrt{3}/2 \end{bmatrix}$$
(3.19)

For the 2-phase machine we want to retain the structure of the equations for flux linkages and write

$$\Psi_{\alpha\beta\circ\mathbf{f}} = L_{\alpha\beta\circ\mathbf{f}} I_{\alpha\beta\circ\mathbf{f}} \tag{3.20}$$

Substituting (3.18), (3.2) and (3.16) into (3.20) and multiplication with  $C_1^{\rm T}$  from the left yields:

$$L_{rstf} = C_1^T L_{\alpha\beta\circ f} C_1$$
 (3.21)

The elements of the inductance matrix  $L_{\alpha\beta of}$  will in general depend on  $\gamma$ . For induced voltages in the 3-phase we have (3.7)

$$V_{rstf} = R_{rstf}I_{rstf} + p\Psi_{rstf}$$
 (3.75)

Substitution of (3.7) into (3.17) gives

$$V_{\alpha\beta\circ f} = C_1 R_{rstf} I_{rstf} + C_1 P_{rstf}^{\Psi}$$
 (3.22)

Using (3.16) and (3.18) to eliminate  $I_{rstf}$  and  $\Psi_{rstf}$  in (3.22) we obtain:

$$V_{\alpha\beta\circ f} = C_1 R_{rstf} C_1^T I_{\alpha\beta\circ f} + C_1 P C_1^T \Psi_{\alpha\beta\circ f}$$

Observing that  $C_1$  does not depend on t we find

$$V_{\alpha\beta\circ f} = R_{\alpha\beta\circ f} I_{\alpha\beta\circ f} + p\Psi_{\alpha\beta\circ f}$$
 (3.23)

where

$$R_{\alpha\beta\delta f} = C_1 R_{rstf} C_1^T$$
 (3.24)

From (3.3) to (3.6) we have

$$R_{rstf} = \begin{bmatrix} r_{a} & 0 & 0 & 0 \\ 0 & r_{a} & 0 & 0 \\ 0 & 0 & r_{a} & 0 \\ 0 & 0 & 0 & r_{f} \end{bmatrix}$$
 (3.25)

Observing that  $C_1$  and  $R_{\text{rstf}}$  comutate we finally have

$$R_{\alpha\beta\circ f} = R_{rstf}$$
 (3.26)

### 3.3. Transformation from a 2-phase Machine to a dq Machine.

In this section we are concerned with the transformation from a 2-phase machine with fixed coils to a dq-machine with moving coils.

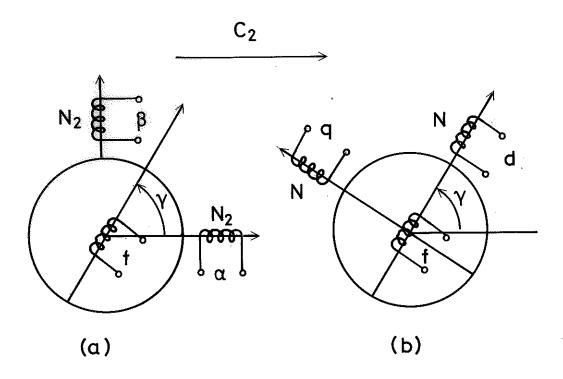


Fig. 3.3 - Transformation from a 2-phase machine to a dq-machine. Original 2-phase machine (a).

Resulting dq-machine (b).

The same requirements are made on this transformation as in the previous section.

Denote the effective number of turns/phase with  $N_2$  and N and divide the mmf into two components on the d-and q-axes respectively. Requirement R 3.1) now gives:

$$Ni_d = N_2(i_\alpha \cos \gamma + i_\beta \sin \gamma)$$

$$Ni_q = N_2(-i_\alpha \sin \gamma + i_\beta \cos \gamma)$$

or in matrix form

$$\begin{bmatrix} \mathbf{i}_{d} \\ \mathbf{i}_{q} \\ \mathbf{i}_{o} \\ \mathbf{i}_{f} \end{bmatrix} = \begin{bmatrix} N_{2}/N \cos \gamma & N_{2}/N \sin \gamma & 0 & 0 \\ -N_{2}/N \sin \gamma & N_{2}/N \cos \gamma & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} \mathbf{i}_{\alpha} \\ \mathbf{i}_{\beta} \\ \mathbf{i}_{o} \\ \mathbf{i}_{f} \end{bmatrix}$$
(3.27)

Requirements R 3.2) and R 3.3) imply, after similar algebra as in the previous section, that:

$$C_2^T C_2 = I \tag{3.28}$$

Condition (3.28) implies

$$N_2 = N$$

Summing up we find that the linear transformation  $C_2$  is given by

$$I_{dqof} = C_2 I_{\alpha\beta of}$$
 (3.29)

$$V_{dqof} = C_2 V_{\alpha\beta of}$$
 (3.30)

$$\Psi_{\text{dqof}} = C_2 \Psi_{\alpha \beta \text{of}} \tag{3.31}$$

where

$$C_{2} = \begin{bmatrix} \cos \gamma & \sin \gamma & 0 & 0 \\ -\sin \gamma & \cos \gamma & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$
 (3.32)

For the dq-machine we write the flux linkage equation

$$\Psi_{\text{dqof}} = L_{\text{dqof}} I_{\text{dqof}}$$
 (3.33)

and postulate that  $L_{\mbox{dqof}}$  shall be independent of  $\gamma$ . The value of the self-inductance of the d-axis winding can be different from the value of the self-inductance of the q-axis winding.

We write the inductance matrix L doof as

$$L_{dqof} = \begin{bmatrix} k+k_2 & 0 & 0 & k_1 \\ 0 & k-k_2 & 0 & 0 \\ 0 & 0 & k_0 & 0 \\ k_1 & 0 & 0 & k_f \end{bmatrix}$$
 (3.34)

where  $\ell$  can be interpreted as the mean value of the self-inductance of an armature winding. The inductance  $\ell_2$  can be interpreted as a variation in self-inductance of an armature winding. The self-inductance of a stator winding has its maximum value when the polar axis of the rotor coincides with the magnetic axis of the stator winding. The minimum value is taken on when the interpolar axis coincides with the magnetic axis of the stator winding. The d- and q-axis windings are orthogonal, which motivates that both  $[L_{dqof}]_{12}$  and  $[L_{dqof}]_{21}$  are zero. The stator leakage inductance is not coupled with any other in-

ductance, which motivates the off-diagonal zeroes in the third row and the third column. The q-axis and field windings are orthogonal, which motivates that both  $[L_{dqof}]_{24}$  and  $[L_{dqof}]_{42}$  are zero.

Substituting (3.20) and (3.29) into (3.33) and multiplication with  $C_2^{\rm T}$  from the left yields:

$$L_{\alpha\beta\circ f} = C_2^T L_{dq\circ f} C_2$$
 (3.35)

which after substitution of (3.32) into (3.35) gives:

$$L_{\alpha\beta\circf} = \begin{bmatrix} \ell + \ell_{2} \cos 2\gamma & \ell_{2} \sin 2\gamma & 0 & \ell_{1} \cos \gamma \\ \ell_{2} \sin 2\gamma & \ell - \ell_{2} \cos 2\gamma & 0 & \ell_{1} \sin \gamma \\ 0 & 0 & \ell_{0} & 0 \\ \ell_{1} \cos \gamma & \ell_{1} \sin \gamma & 0 & \ell_{f} \end{bmatrix}$$
(3.36)

To derive an expression for  $L_{rstf}$  we use equation (3.21)

$$L_{rstf} = C_1^T L_{\alpha\beta\alpha f} C_1 \tag{3.21}$$

Substitution of (3.36) into (3.21) now yields:

$$L_{rstf} = \begin{bmatrix} l & 11 & l & 12 & l & 13 & l & 14 \\ l & 21 & l & 22 & l & 23 & l & 24 \\ l & 31 & l & 32 & l & 33 & l & 34 \\ l & 41 & l & 42 & l & 43 & l & 44 \end{bmatrix}$$
(3.37)

where

$$\begin{array}{l} {\ell_{11}} = {\ell_{0}}/2 + {\ell_{1}} + {\ell_{2}} \cos 2\gamma \\ \\ {\ell_{12}} = {\ell_{21}} = -{\ell/2} + {\ell_{2}} \cos(2\gamma - 2\pi/3) \\ \\ {\ell_{13}} = {\ell_{31}} = -{\ell/2} + {\ell_{2}} \cos(2\gamma + 2\pi/3) \\ \\ {\ell_{14}} = {\ell_{41}} = \sqrt{3/2} {\ell_{1}} \cos \gamma \\ \\ {\ell_{22}} = {\ell_{0}}/2 + {\ell_{1}} + {\ell_{2}} \cos(2\gamma + 2\pi/3) \\ \\ {\ell_{23}} = {\ell_{32}} = -{\ell/2} + {\ell_{2}} \cos(2\gamma + 2\pi/3) \\ \\ {\ell_{24}} = {\ell_{42}} = \sqrt{3/2} {\ell_{1}} \cos(\gamma - 2\pi/3) \\ \\ {\ell_{33}} = {\ell_{0}}/2 + {\ell_{1}} + {\ell_{2}} \cos(2\gamma - 2\pi/3) \\ \\ \\ {\ell_{34}} = {\ell_{43}} = \sqrt{3/2} {\ell_{1}} \cos(\gamma + 2\pi/3) \\ \\ \\ {\ell_{111}} = 3/2 + {\ell_{1}} \end{array}$$

We now observe that assumption A 3.1) and assumption A 3.2) allow  $\rm L_{dq\,o\,f}$  to be a constant matrix.

To derive an expression for induced voltages in the dq machine we use equation (3.23)

$$V_{\alpha\beta\circ f} = R_{\alpha\beta\circ f} I_{\alpha\beta\circ f} + p\Psi_{\alpha\beta\circ f}$$
 (3.23)

A substitution of (3.29) and (3.31) into (3.23) yields after multiplication with  $C_2$  from the left

$$V_{dqof} = C_2 R_{\alpha\beta\circ f} C_2^T I_{dq\circ f} + C_2 P (C_2^T \Psi_{dq\circ f})$$
 (3.38)

Taking derivative of the second term in (3.38) gives

$$V_{dq \circ f} = R_{dq \circ f} I_{dq \circ f} + C_{2} P(C_{2}^{T}) \Psi_{dq \circ f} + P \Psi_{dq \circ f}$$
 (3.39)

where R<sub>dqof</sub> is given by

$$R_{dqof} = C_2 R_{\alpha\beta\circ f} C_2^{T}$$
 (3.40)

Using (3.26) we find

$$R_{dqof} = C_2 R_{rstf} C_2^{T}$$
 (3.41)

Observing that  $\rm C_2$  and  $\rm R_{rstf}$  commutate we have the following expression for  $\rm R_{dqo\,f}$ 

$$R_{dqof} = R_{rstf}$$
 (3.42)

Introducing

(3.39) can be written

$$V_{dqof} = R_{dqof}I_{dqof} + W_{dqof} + p_{dqof}$$
 (3.44)

The first term in equation (3.44) represents voltage drop across the armature resistance. The second term represents the speed voltages and the third term represents transformer voltages.

#### 3.4. The Air-Gap Torque.

To derive an expression for the air-gap torque we apply the principle of conservation of energy, which can be formulated

$$\frac{dE}{dt} = \frac{dE_e}{dt} + \frac{dE_m}{dt} = P_m + P_n - P_{\ell r} - P_{\ell m} - P_{\ell d} \qquad (3.45)$$

where

 $E_{_{\mathbf{e}}}$  = energy stored in the magnetic circuits

 $E_{m}$  = energy stored in the rotating masses

 $P_{m}$  = power delivered from the prime mover

P<sub>n</sub> = power delivered from the network

 $P_{\ell r}$  = power losses in the resistances

 $P_{lm}$  = mechanical power losses

 $P_{\ell d}$  = power losses in the damping winding

The energy stored in the magnetic circuits can be written

$$E_{e} = \frac{1}{2} I_{dqof}^{T} L_{dqof} I_{dqof}$$
 (3.46)

The energy stored in the rotating masses is given by

$$E_{m} = \frac{1}{2} J \omega^{2}$$
 (3.47)

where

J = moment of inertia of the combined turbine generator

 $\omega$  = angular velocity of rotor

The electrical power delivered from the network to the generator is given by the expression

$$P_{n} = I_{dqof}^{T} V_{dqof}$$
 (3.48)

Substituting (3.44) into (3.48) gives

$$P_{n} = I_{dqof}^{T} R_{dqof} I_{dqof} + I_{dqof}^{T} W_{dqof} + I_{dqof}^{T} P_{dqof}$$

$$+ I_{dqof}^{T} P_{dqof}$$
(3.49)

The power losses in the resistances is given by

$$P_{lr} = I_{dqof}^{T} R_{dqof} I_{dqof}$$
 (3.50)

Finally we assume that the power losses in the damping windings can be written

$$P_{\ell d} = D_1 \omega (\omega - \omega_0)$$
 (3.51)

and the mechanical power losses can be written

$$P_{\ell m} = D_2 \omega^2 \tag{3.52}$$

Taking derivatives of (3.47) and (3.48) now gives

$$\frac{dE_{m}}{dt} = J_{\omega} \frac{d\omega}{dt}$$
 (3.53)

$$\frac{dE_{e}}{dt} = I_{dqof}^{T} \frac{d}{dt} (L_{dqof}I_{dqof}) = I_{dqof}^{T}P^{\psi}dqof$$
 (3.54)

Substitution of (3.49) to (3.54) into (3.45) and rearranging the terms yields

$$J \frac{d\omega}{dt} = P_{\rm m}/\omega - M_{\rm e} - M_{\rm d}$$
 (3.55)

where the air-gap torque  $\mathbf{M}_{\mathbf{e}}$  is given by

$$M_{e} = -I_{dq \circ f}^{T} W \psi_{dq \circ f} / \omega = \psi_{q} i_{d} - \psi_{d} i_{q}$$
 (3.56)

The damping torque  $\mathbf{M}_{\mathbf{d}}$  is given by

$$M_{d} = D_{1}(\omega - \omega_{0}) + D_{2}\omega$$
 (3.57)

Equations (3.55), (3.56) and (3.57) will be used in the following sections.

### 3.5. Linearized Equations for the Synchronous Generator.

The nonlinear equations for the synchronous machine will now be linearized. To avoid a lot of negative numerical values of the generator currents we will also change the sign conventions for  $i_d$ ,  $i_q$  and  $i_o$ . Motor references were previously used for all circuits. We will now use generator references for dand q-axis windings as well as for the zero sequence winding.



Fig. 3.42 - Sign conventions. Motor references (a) and generator references (b).

Motor references means that applied voltage and current into the winding are positive.

Generator references, on the other hand, means that generated voltage and current out of the winding are positive.

In Section 3.3 we derived expressions for the <u>flux</u> <u>linkages</u>

$$\Psi_{\text{dqof}} = L_{\text{dqof}} I_{\text{dqof}} \tag{3.33}$$

$$L_{dqof} = \begin{pmatrix} \ell + \ell_2 & 0 & 0 & \ell_1 \\ 0 & \ell - \ell_2 & 0 & 0 \\ 0 & 0 & \ell_0 & 0 \\ \ell_1 & 0 & 0 & \ell_f \end{pmatrix}$$
 (3.347)

After change of sign conventions we can write

$$\psi_{f} = L_{f}i_{f} - L_{af}i_{d} \qquad (3.58)$$

$$\psi_{d} = L_{af}i_{f} - L_{d}i_{d}$$
 (3.59)

$$\psi_{\mathbf{q}} = - \mathbf{L}_{\mathbf{q}} \mathbf{i}_{\mathbf{q}} \tag{3.60}$$

where we have introduced

$$L_{f} = \ell_{f}$$

$$L_d = l + l_2$$

$$L_{G} = l - l_{2}$$

As the equations (3.58) to (3.60) already are linear they are immediately valid for small deviations and we have

$$\begin{bmatrix} \delta \psi_{f} \\ \delta \psi_{d} \\ \delta \psi_{q} \end{bmatrix} = \begin{bmatrix} L_{f} & -L_{af} & 0 \\ L_{af} & -L_{d} & 0 \\ 0 & 0 & -L_{q} \end{bmatrix} \begin{bmatrix} \delta i_{f} \\ \delta i_{d} \\ \delta i_{q} \end{bmatrix}$$
 (3.61)

where  $\delta x = x - x_0$ . Equation (3.61) can be written symbolically as

$$\delta \Psi = L \delta I \tag{3.62}$$

which after multiplication with  $\omega_{o}$  yields

$$\delta \omega_{o} \Psi = X \delta I \tag{3.63}$$

where

$$X = \omega_0 L \tag{3.64}$$

In Section 3.3 we also derived expressions for induced voltages in the dq machine.

$$V_{dqof} = R_{dqof}I_{dqof} + W_{dqof} + P_{dqof}$$
 (3.44')

After change of sign conventions (3.43') and (3.44') can be written in component form

$$v_f = p\psi_f + r_f i_f \tag{3.65}$$

$$v_{d} = p\psi_{d} - r_{a}i_{d} - \omega\psi_{d}$$
 (3.66)

$$v_{q} = p\psi_{q} - r_{a}i_{q} + \omega\psi_{d}$$
 (3.67)

After linearization (3.65) to (3.67) become

$$\delta v_f = p \delta \psi_f + r_f \delta i_f \qquad (3.68)$$

$$\delta v_{d} = p \delta \psi_{d} - r_{a} \delta i_{d} - \omega_{o} \delta \psi_{q} - \psi_{q} \delta \omega \qquad (3.69)$$

$$\delta v_{q} = p \delta \psi_{q} - r_{a} \delta i_{q} + \omega_{o} \delta \psi_{d} + \psi_{d} \delta \omega \qquad (3.70)$$

where the angular velocity  $\omega = d\gamma/dt$ .

In Section 3.4 we derived the following expression for the air-gap torque

$$M_e = \psi_a i_d - \psi_d i_a$$
 (3.56')

After change of sign conventions (3.56') can be written

$$M_{e} = \psi_{d} i_{q} - \psi_{q} i_{d}$$
 (3.71)

which after linearization becomes

$$\delta M_{e} = i_{q} \delta \psi_{d} - i_{d} \delta \psi_{q} - \psi_{q} \delta i_{d} + \psi_{d} \delta i_{q}$$
 (3.72)

#### 3.6. Basic Equations for the Exciter.

The exciter system of each generator is assumed to be described by a first order linear system with time-constant  $\mathbf{T}_{\mathbf{e}}$ .

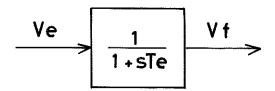


Fig. 3.5 - Block diagram for the exciter.

The differential equation describing the exciter is obtained from Fig. 3.5

$$pv_f = (-v_f + v_e)/T_e$$
 (3.73)

Since equation (3.73) already is linear it is also valid for small deviations from an equilibrium point and we have

$$p\delta v_{f} = (-\delta v_{f} + \delta v_{e})/T_{e}$$
 (3.74)

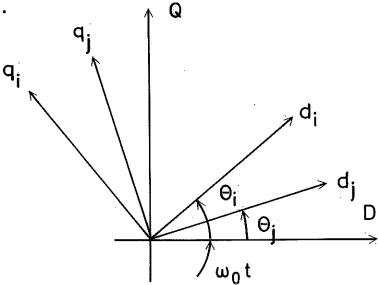
#### 4. TRANSMISSION NETWORK.

In the previous sections we derived equations for the individual generators and used the polar and the interpolar axes of the rotor as reference frames for the electric quantities. These axes do not in general coincide with corresponding axes of another generator. In this section we choose a common frame of references for the electric quantities. We also derive a transformation from rotor-based to networkbased quantities.

#### 4.1. Selection of Angular References.

The equations for each generator are expressed with reference to pairs of axes (d,q) which rotate in synchronism with the rotor of the generators. On the other hand, the equations for the connecting network refer to axes (D,Q) rotating at constant speed ( $\omega_0$ ). In steady-state these axes rotate at the same speed.

The angular displacements, defined in Fig. 4.1, can be obtained from the solution of the load-flow problem.



<u>Fig. 4.1</u> - Angular relationships between network and synchronous machine reference axes.

The choice of common reference frame is not unique. One reasonable choice is that (D,Q) coincide with  $(d_{\ell},q_{\ell})$ , the reference frame of the largest generator, in steady-state.

Under transient conditions the angles  $\theta_i$  will vary as the machine speeds vary. The angles  $\theta_i$  are state-variables and  $i_d$  and  $i_q$  are linear combinations of state-variables but  $v_d$  and  $v_q$  are needed for the computation of  $p\delta\psi_d$  and  $p\delta\psi_q$ . Therefore it is necessary to have an expression for  $\delta v_d$  and  $\delta v_q$  in  $\delta\theta$ ,  $\delta i_d$  and  $\delta i_q$ .

#### 4.2. Transformation of Network Equations.

The transformation relating rotor-based voltages to network-based voltages is given in [10]

$$\begin{bmatrix} v_{d1} \\ v_{q1} \\ \vdots \\ v_{dn} \\ v_{qn} \end{bmatrix} = \begin{bmatrix} \cos \theta_1 & \sin \theta_1 & \cdots & 0 & 0 \\ -\sin \theta_1 & \cos \theta_1 & \cdots & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & \cdots & \cos \theta_n & \sin \theta_n \\ 0 & 0 & \cdots & -\sin \theta_n & \cos \theta_n \end{bmatrix} \begin{bmatrix} v_{D1} \\ v_{Q1} \\ \vdots \\ v_{Dn} \\ v_{Qn} \end{bmatrix}$$

$$(4.1)$$

For the present approach we reorder the equations in (4.1) to obtain

$$\begin{bmatrix} v_{d1} \\ \vdots \\ v_{dn} \\ v_{q1} \\ \vdots \\ v_{qn} \end{bmatrix} = \begin{bmatrix} \cos \theta_1 & \sin \theta_1 & 0 \\ \vdots & 0 & \vdots \\ \cos \theta_n & \sin \theta_n \\ \cos \theta_1 & v_{Dn} \\ \cos \theta_1 & \cos \theta_1 \\ \vdots & \vdots \\ \cos \theta_n & \cos \theta_1 \\ \vdots & \vdots \\ \cos \theta_n & \cos \theta_n \end{bmatrix} \begin{bmatrix} v_{D1} \\ \vdots \\ v_{Dn} \\ v_{Q1} \\ \vdots \\ v_{Qn} \end{bmatrix}$$

Equation (4.2) can be written

$$\begin{bmatrix} v_{d} \\ v_{q} \end{bmatrix} = \begin{bmatrix} c & s \\ -s & c \end{bmatrix} \begin{bmatrix} v_{D} \\ v_{Q} \end{bmatrix}$$
 (4.3)

where

$$v_d = (v_{d1}, v_{d2}, ..., v_{dn})^T$$
 (4.4)

$$V_q = (v_{q1}, v_{q2}, ..., v_{qn})^T$$
 (4.5)

$$V_{D} = (V_{D1}, V_{D2}, ..., V_{Dn})^{T}$$
 (4.6)

$$V_Q = (v_{Q1}, v_{Q2}, ..., v_{Qn})^T$$
 (4.7)

$$C = diag(\cos \theta_1, \cos \theta_2, \dots, \cos \theta_n)$$
 (4.8)

$$S = diag(\sin \theta_1, \sin \theta_2, ..., \sin \theta_n)$$
 (4.9)

We also need a transformation from rotor-based currents to network-based currents. We require that the power  $I^{T}V$  shall be invariant under the transformation. To derive the transformation matrix we rewrite (4.3)

$$V_{M} = TV_{N}$$
 (4.10)

where

$$v_{M} = (v_{d}^{T}, v_{q}^{T})^{T}$$

$$v_{N} = (v_{D}^{T}, v_{Q}^{T})^{T}$$

$$T = \begin{bmatrix} C & \bar{S} \\ -S & C \end{bmatrix}$$

The power invariance requires

$$I_N^T V_N = I_M^T V_M = I_M^T I_N^T V_N$$

which implies

$$I_{N} = T^{T}I_{M}$$
 (4.11)

where

$$I_{N} = (I_{D}^{T}, I_{Q}^{T})^{T}$$

$$\mathbf{I}_{\mathbf{M}} = (\mathbf{I}_{\mathbf{d}}^{\mathbf{T}}, \mathbf{I}_{\mathbf{q}}^{\mathbf{T}})^{\mathbf{T}}$$

In Section 2 we stated that the network could be described by the nodal admittance matrix in (2.3)

$$\begin{bmatrix} I_D \\ I_Q \end{bmatrix} = \begin{bmatrix} G_N & -B_N \\ B_N & G_N \end{bmatrix} \begin{bmatrix} V_D \\ V_Q \end{bmatrix}$$
 (2.37)

The nodal admittance matrix is always nonsingular, making it possible to write

$$\begin{bmatrix} V_{D} \\ V_{Q} \end{bmatrix} = \begin{bmatrix} R_{N} & -X_{N} \\ X_{N} & R_{N} \end{bmatrix} \begin{bmatrix} I_{D} \\ I_{Q} \end{bmatrix}$$
 (4.12)

Substitution of (4.11) and (4.12) into (4.3) now yields

$$\begin{bmatrix} V_{d} \\ V_{q} \end{bmatrix} = \begin{bmatrix} C & S \\ -S & C \end{bmatrix} \begin{bmatrix} R_{N} & -X_{N} \\ X_{N} & R_{N} \end{bmatrix} \begin{bmatrix} C & -S \\ S & C \end{bmatrix} \begin{bmatrix} I_{d} \\ I_{q} \end{bmatrix}$$
(4.13)

which can be written

$$\begin{bmatrix} V_{d} \\ V_{q} \end{bmatrix} = \begin{bmatrix} R_{m} & -X_{m} \\ X_{m} & R_{m} \end{bmatrix} \begin{bmatrix} I_{d} \\ I_{q} \end{bmatrix}$$
(4.14)

where

$$R_{m} = CR_{N}C + SR_{N}S + SX_{N}C - CX_{N}S$$
 (4.15)

$$X_{m} = CX_{N}C + SX_{N}S + CR_{N}S - SR_{N}C$$
 (4.16)

The linearized version of (4.14) can be written

$$\begin{bmatrix} \delta V_{d} \\ \delta V_{q} \end{bmatrix} = \begin{bmatrix} R_{m} & -X_{m} \\ X_{m} & R_{m} \end{bmatrix} \begin{bmatrix} \delta I_{d} \\ \delta I_{q} \end{bmatrix} + \begin{bmatrix} E_{d} \\ E_{q} \end{bmatrix} \delta / \Theta$$
 (4.17)

where

$$E_{d} = \frac{\partial}{\partial \theta} \left( R_{m} I_{d} - X_{m} I_{q} \right) \tag{4.18}$$

and

$$E_{q} = \frac{\partial}{\partial \Theta} \left( X_{m} I_{d} + R_{m} I_{q} \right) \tag{4.19}$$

#### 5. PRIME MOVERS.

The fundamental torque balance equation was derived in Section 3.4

$$Jp\omega = P_{m}/\omega - M_{e} - M_{d}$$
 (3.55)

where

J = moment of inertia of the combined turbine generator

 $\omega$  = angular velocity of the rotor

 $P_{\rm m}$  = mechanical power delivered from the prime mover

M<sub>e</sub> = air-gap torque

 $M_d$  = damping torque

Assuming

$$M_d = D_1(\omega - \omega_0) + D_2\omega$$

we find

$$Jp\delta\omega = \delta P_{\rm m}/\omega_{\rm o} - \delta M_{\rm e} - D\delta\omega \qquad (5.1)$$

where

$$D = D_1 + D_2 + P_m/\omega_0$$
 (5.2)

To obtain expressions for  $\delta P_{m}$  it is necessary to investigate the different types of prime movers.

#### 5.1. Hydro Turbines.

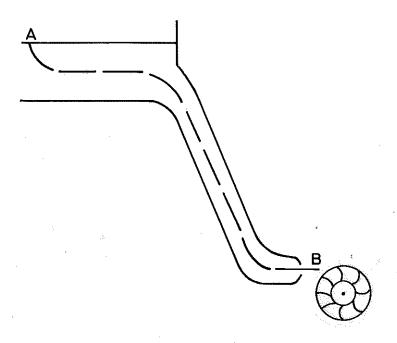


Fig. 5.1 - Simplified diagram of a hydro turbine.

Following [13] we have for the hydro turbine in Fig. 5.1

$$P_{m} = \frac{1}{2} \rho v_{out}^{2} q = \frac{1}{2} \rho a v_{out}^{3}$$
 (5.3)

which essentially states that all potential energy is converted to kinetic energy and that all kinetic energy is available as output power from the hydro turbine.

In equation (5.3)

P<sub>m</sub> - mechanical output power

q - flow of water

ρ - density of water

 $v_{\text{out}}$  - velocity of water at outlet

a - outlet area

In practice it is observed [15] that the efficiency is depending on the angular velocity of the hydro turbine. This means that (5.3) has to be written

$$P_{m} = \frac{1}{2} \eta(\omega) \rho \cdot av_{OUT}^{3} \qquad (5.3)$$

where  $\eta(\omega)$  is a speed dependent efficiency. Bernoulli's theorem yields further

$$Lpv_t + \frac{1}{2}v_{out}^2 - gh = 0$$
 (5.4)

where

 $\mathbf{v_{+}}$  - velocity of water in the dash-tube

L - length of the dash-tube

g - constant of gravity

h - water head

The dash-tube is assumed to have the constant area A. The equation of continuity implies

$$av_{out} = Av_{t}$$
 (5.5)

A substitution of (5.5) into (5.6) yields

$$pv_{\pm} = gh/L - (Av_{\pm}/a)^2/2L$$
 (5.6)

Introducing the maximum steady-state velocity of the water in the dash-tube

$$v_{\text{tmax}} = \sqrt{2gh} a_{\text{max}}/A$$
 (5.7)

and the state-variable

$$z_{p} = v_{t}/v_{tmax}$$
 (5.8)

we obtain

$$pz_p = \frac{A\sqrt{2gh}}{2La_{max}} (1 - z_p^2 a_{max}^2 / a^2)$$
 (5.9)

Substituting

$$u_{t} = a/a_{max}$$
 (5.10)

$$T\omega = La_{max}/A\sqrt{2gh}$$
 (5.11)

Equations (5.3) and (5.9) are transferred into

$$pz_p = (1 - z_p^2/u_t^2)/T_w$$
 (5.12)

$$P_{m} = P_{max}(\omega)z_{p}^{3}/u_{t}^{2}$$
 (5.13)

Linearization of (5.12) and (5.13) finally yields

$$pz_{p} = (-\delta z_{p} + \delta u_{t})/T_{w}$$
 (5.14)

$$\delta P_{m} = P_{max}(\omega_{o})(3\delta z_{p} - 2\delta u_{t}) + \frac{\partial P_{max}}{\partial \omega}(\omega_{o})\delta \omega \qquad (5.15)$$

If we include

$$\frac{\partial P_{\text{max}}}{\partial \omega} (\omega_{o}) \delta \omega$$

in the general damping term in (5.2) we can write (5.14) and (5.15) as

$$p\delta z_{p} = a\delta z_{p} + b\delta u_{t}$$
 (5.16)

$$\delta P_{\rm m} = c \delta z_{\rm p} + d \delta u_{\rm t}$$
 (5.17)

#### 5.2. Boilers and Steam Turbines.

Aström and Eklund [14] have shown that a reasonable accurate and low order dynamical model of a boiler and steam turbine unit is given by

$$\frac{dp}{dt} = -\alpha_1(u_1p^{5/8} - \alpha_5) + \alpha_2u_2 - \alpha_3u_3$$
 (5.18)

where

p - steam pressure

u<sub>1</sub> - steam valve setting

u<sub>2</sub> - fuel flow

u<sub>3</sub> - feedwater flow

The mechanical output power from the turbine is given by

$$P_{m} = \alpha_{4}(u_{1}p^{5/8} - \alpha_{5})$$
 (5.19)

The model (5.18) is essentially an energy balance equation and it is assumed that the stored energy in the boiler mainly depends on the steam pressure.

The first term in the right member of (5.18) represents the energy in the steam delivered from the boiler to the turbine. The second term represents energy supplied from the fuel while the last term represents the cooling effect of the feedwater. The output power is proportional to the energy flow from the boiler to the turbine, which implies that the boiler steam-turbine plant has constant efficiency.

In this application we are not allowed to vary  $u_3$ , the feedwater flow, independent of the state of the boiler. Instead we assume that the boiler is equipped with a feedwater regulator, which provides the boiler with feedwater flow proportional to the steam flow. The steam flow is given by

$$q \sim u_1 \sqrt{p}$$

We also introduce the following normalized variables into (5.18) and (5.19)

$$z_{p} = p/p_{max}$$
 (5.20)

$$u_t = u_1/u_{1max}$$
 (5.21)

$$u_f = u_2/u_{2max}$$
 (5.22)

and prescribe that the differential equation (5.18) shall have a stationary point at

 $p = p_{max}$ 

 $u_1 = u_{1max}$ 

 $u_2 = u_{2max}$ 

 $u_3 = u_{3max}$ 

 $P_b = P_{max}$ 

Equations (5.18) and (5.19) can then be written

$$pz_{p} = \left\{-[(1+\alpha)u_{t}z_{p}^{5/8} - \alpha] + (1+\beta)u_{f} - \beta u_{t}z_{p}^{1/2}\right\}/T_{b}$$
(5.23)

$$P_{\rm m} = P_{\rm max}[(1+\alpha)u_{\rm t}z_{\rm p}^{5/8} - \alpha]$$
 (5.24)

The linearized version of equations (5.23) and (5.24) becomes

$$p\delta z_p = a\delta z_p + b_t \delta u_t + b_f \delta \omega_f$$
 (5.25)

$$\delta P_{m} = c \delta z_{p} + d_{t} \delta u_{t}$$
 (5.26)

#### 6. CONSTRUCTION OF SYSTEM MATRICES.

In the previous sections we derived the equations for one machine at a time but we are now going to derive the differential equations for n interconnected generating plants. To simplify the derivations and to improve the clarity we partition the state and the input vector into subvectors. The system matrices A and B in (1.1) are similarly partitioned into submatrices. In this section we are going to derive the differential equations for the subvectors one by one.

The state vector x is partitioned into seven subvectors in the following manner

$$x^{T} = (x_{1}^{T}, x_{2}^{T}, x_{3}^{T}, x_{4}^{T}, x_{5}^{T}, x_{6}^{T}, x_{7}^{T})$$
 (6.1)

where

$$\mathbf{x}_{1}^{\mathrm{T}} = (\delta \Theta_{1}, \delta \Theta_{2}, \dots, \delta \Theta_{n}) \tag{6.2}$$

$$\mathbf{x}_{2}^{\mathrm{T}} = (\delta \omega_{1}, \delta \omega_{2}, \dots, \delta \omega_{p}) \tag{6.3}$$

$$\mathbf{x}_{3}^{\mathrm{T}} = (\delta \omega_{0} \psi_{f1}, \delta \omega_{0} \psi_{f2}, \dots, \delta \omega_{0} \psi_{fn})$$
 (6.4)

$$\mathbf{x}_{4}^{T} = (\delta \omega_{0} \psi_{d1}, \delta \omega_{0} \psi_{d2}, \dots, \delta \omega_{0} \psi_{dn})$$
 (6.5)

$$\mathbf{x}_{5}^{\mathrm{T}} = (\delta \omega_{0} \psi_{q1}, \delta \omega_{0} \psi_{q2}, \dots, \delta \omega_{0} \psi_{qn}) \tag{6.6}$$

$$x_6^T = (\delta e_{f1}, \delta e_{f2}, ..., \delta e_{fn})$$
 (6.7)

$$x_7^T = (\delta z_{p1}, \delta z_{p2}, \dots, \delta z_{pn})$$
 (6.8)

where  $e_{fi} = x_{afi}v_{fi}/r_{fi}$  and  $z_{pi}$  is the prime mover state-variable for the i:th plant ( $z_h$  or  $z_b$ ).

The input vector u is partitioned into 3 subvectors in the following manner

$$u^{T} = (u_{1}^{T}, u_{2}^{T}, u_{3}^{T})$$
 (6.9)

where

$$u_1^T = (\delta u_{e1}, \delta u_{e2}, ..., \delta u_{en})$$
 (6.10)

$$u_2^T = (\delta u_{t1}, \delta u_{t2}, \dots, \delta u_{tn})$$
 (6.11)

$$u_3^T = (\delta u_{f1}, \delta u_{f2}, ..., \delta u_{fn})$$
 (6.12)

Here uei denotes the input signal to exciter i, uti denotes the first input signal to prime mover No. i (gate opening or steam valve setting) and ufi denotes the second input signal to prime mover No. i (fuel flow for steam plants).

In a similar way the system matrices A and B are partitioned into submatrices and we can write (1.1) as

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} = \begin{bmatrix} A_{11} & A_{12} & A_{13} & A_{14} & A_{15} & A_{16} & A_{17} \\ A_{21} & A_{22} & A_{23} & A_{24} & A_{25} & A_{26} & A_{27} \\ A_{31} & A_{32} & A_{33} & A_{34} & A_{35} & A_{36} & A_{37} \\ A_{41} & A_{42} & A_{43} & A_{44} & A_{45} & A_{46} & A_{47} \\ A_{51} & A_{52} & A_{53} & A_{54} & A_{55} & A_{56} & A_{57} \\ A_{61} & A_{62} & A_{63} & A_{64} & A_{65} & A_{66} & A_{67} \\ A_{71} & A_{72} & A_{73} & A_{74} & A_{75} & A_{76} & A_{77} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \\ x_5 \\ x_6 \\ x_7 \end{bmatrix}$$

$$\begin{bmatrix} B_{11} & B_{12} & B_{13} \\ B_{21} & B_{22} & B_{23} \\ B_{31} & B_{32} & B_{33} \\ B_{41} & B_{42} & B_{43} \\ B_{51} & B_{52} & B_{53} \\ B_{61} & B_{62} & B_{63} \\ B_{71} & B_{72} & B_{73} \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \\ u_3 \end{bmatrix}$$

We are now going to derive expressions for the submatrices one row at a time.

### 6.1. Differential Equations for Rotor Angles (x<sub>1</sub>).

The rotor angle is defined by

$$\theta_{i}(t) = \int_{0}^{t} \omega_{i}(s) ds - \omega_{o}t + \theta_{io}$$

which immediately gives

$$p\delta\Theta_{i} = \omega_{i}(t) - \omega_{o} = \delta\omega_{i}$$
 (6.13)

or symbolically

$$p\delta \Theta = A_{12}\delta\Omega \tag{6.14}$$

where

$$\delta \Theta = (\delta \Theta_1, \delta \Theta_2, \dots, \delta \Theta_n)^T$$
 (6.15)

$$\delta\Omega = (\delta\omega_1, \delta\omega_2, \ldots, \delta\omega_n)^{\mathrm{T}}$$
 (6.16)

$$A_{12} = I$$
 (6.17)

# 6.2. Differential Equations for Rotor Angular Velocities $(x_2)$ .

The torque balance equation for the rotor was derived in Section 5.

$$p\delta\omega_{i} = (\delta P_{mi}/\omega_{o} - \delta M_{ei} - D_{i}\delta\omega_{i})/J_{i}$$
 (5.1')

where  $\delta P_{mi}$  is the mechanical input power from the prime mover given by (5.17) or (5.26)

$$\delta P_{mi} = c_i \delta z_{pi} + d_i \delta u_{ti}$$
 (5.17)

$$\delta P_{mi} = c_{i} \delta z_{pi} + d_{i} \delta u_{ti}$$
 (5.26')

 $\delta M_{ei}$  is the air-gap torque given by (3.72)

$$\delta M_{ei} = i_{qi} \delta \psi_{di} - i_{di} \delta \psi_{qi} - \psi_{qi} \delta i_{di} + \psi_{di} \delta i_{qi}$$
 (3.72')

Collecting the torque balance equations for all rotor we have

$$P^{\delta\Omega} = A_{22}^{\delta\Omega} + G_{q}^{\delta I}_{d} + G_{d}^{\delta I}_{q} + H_{q}^{\delta\omega} + H_{q}^{\delta\omega} + H_{d}^{\delta\omega} + H_{d}^{\omega} + H_{d}^{\delta\omega} + H_{d}^{\delta\omega} + H_{d}^{\delta\omega} + H_{d}^{\delta\omega} + H_{d}^{\omega} + H_{d}^{\delta\omega} + H_{d}^{\delta\omega} + H_{d}^{\delta\omega} + H_{d}^{\delta\omega} + H_{d}^{\omega} + H_{d}^{\delta\omega} + H_{d}^{\delta\omega} + H_{d}^{\delta\omega} + H_{d}^{\delta\omega} + H_{d}^{\omega} + H_{d}^{\delta\omega} + H_{d}^{\omega} + H_{d}^{$$

where

$$\delta I_{d} = (\delta i_{d1}, \delta i_{d2}, \dots, \delta i_{dn})^{T}$$
(6.19)

$$\delta I_{q} = (\delta i_{q1}, \delta i_{q2}, \dots, \delta i_{qn})^{T}$$
(6.20)

$$\delta \omega_{o}^{\Psi}_{d} = (\delta \omega_{o} \psi_{d1}, \delta \omega_{o} \psi_{d2}, \dots, \delta \omega_{o} \psi_{dn})^{T}$$
 (6.21)

$$\delta \omega_{o}^{\Psi}_{q} = (\delta \omega_{o}^{\Psi}_{q1}, \delta \omega_{o}^{\Psi}_{q2}, \dots, \delta \omega_{o}^{\Psi}_{qn})^{T}$$
 (6.22)

$$\delta Z_{p} = (\delta z_{p1}, \delta z_{p2}, \dots, \delta z_{pn})^{T}$$
(6.23)

$$\delta U_{t} = (\delta u_{t1}, \delta u_{t2}, \dots, \delta u_{tn})^{T}.$$
 (6.24)

$$A_{22} = diag(-D_1/J_1, -D_2/J_2, ..., -D_n/J_n)$$
 (6.25)

$$G_q = diag(-\psi_{q1}/J_1, -\psi_{q2}/J_2, ..., -\psi_{qn}/J_n)$$
 (6.26)

$$G_d = diag(\psi_{d1}/J_1, \psi_{d2}/J_2, ..., \psi_{dn}/J_n)$$
 (6.27)

$$H_{q} = diag(i_{q1}/\omega_{o}J_{1}, i_{q2}/\omega_{o}J_{2}, ..., i_{qn}/\omega_{o}J_{n})$$
 (6.28)

$$H_d = diag(-i_{d1}/\omega_o J_1, -i_{d2}/\omega_o J_2, \dots, -i_{dn}/\omega_o J_n)$$
 (6.29)

$$A_{27} = diag(c_1/\omega_0 J_1, c_2/\omega_0 J_2, ..., c_n/\omega_0 J_n)$$
 (6.30)

$$B_{22} = diag(d_1/\omega_o J_1, d_2/\omega_o J_2, ..., d_n/\omega_o J_n)$$
 (6.31)

To derive an expression for  $I_d$  and  $I_q$  we use equation (3.63) derived in Section 3.5

$$\delta \omega_{o} \Psi = X \delta I$$
 (3.63<sup>^</sup>)

The matrix X in (3.63') is always nonsingular for physical reasons.

The stored energy in the magnetic circuits can be written  $\textbf{I}^{T}\textbf{XI/}\omega_{_{\Omega}}$  and the stored energy is always positive

if any current differs from zero. Hence we can always express  $\delta I$  in terms of  $\delta \omega_0^{\ \Psi}$ . Equation (3.63') can be written

$$\begin{bmatrix} \delta \omega_{o} \psi_{f} \\ \delta \omega_{o} \psi_{d} \\ \delta \omega_{o} \psi_{q} \end{bmatrix} = \begin{bmatrix} X_{f} & -X_{af} & 0 \\ X_{af} & -X_{d} & 0 \\ 0 & 0 & X_{q} \end{bmatrix} \begin{bmatrix} \delta i_{f} \\ \delta i_{d} \\ \delta i_{q} \end{bmatrix}$$
 (6.32)

where

$$X_{f} = \hat{w}_{o}L_{f} \tag{6.33}$$

$$X_{af} = \omega_{o}L_{af} \tag{6.34}$$

$$X_{d} = \omega_{o}L_{d}$$
 (6.35)

$$X_{q} = \omega_{o}L_{q} \tag{6.36}$$

Knowing that the inverse of X exists and observing the structure of the matrix we have

$$\begin{bmatrix} \delta \mathbf{i}_{\mathbf{f}} \\ \delta \mathbf{i}_{\mathbf{d}} \end{bmatrix} = \begin{bmatrix} y_{\mathbf{f}\mathbf{f}} & y_{\mathbf{f}\mathbf{d}} & 0 \\ y_{\mathbf{d}\mathbf{f}} & y_{\mathbf{d}\mathbf{d}} & 0 \\ 0 & 0 & y_{\mathbf{q}\mathbf{q}} \end{bmatrix} \begin{bmatrix} \delta \omega_{\mathbf{0}} \psi_{\mathbf{f}} \\ \delta \omega_{\mathbf{0}} \psi_{\mathbf{d}} \\ \delta \omega_{\mathbf{0}} \psi_{\mathbf{q}} \end{bmatrix}$$
(6.37)

Now it is possible to form the expressions for  $\delta \mathbf{I}_{f},$   $\delta \mathbf{I}_{d}$  and  $\delta \mathbf{I}_{q}.$ 

$$\delta I_{f} = Y_{ff} \delta \omega_{o}^{\Psi} + Y_{fd} \delta \omega_{o}^{\Psi} d \qquad (6.38)$$

$$\delta I_{d} = Y_{df} \delta \omega_{o}^{\Psi} + Y_{dd} \delta \omega_{o}^{\Psi} d \qquad (6.39)$$

$$\delta I_{q} = Y_{qq} \delta \omega_{o}^{\Psi}_{q} \qquad (6.40)$$

where

$$\delta I_f = (\delta i_{f1}, \delta i_{f2}, \dots, \delta i_{fn})^T$$
(6.41)

$$\delta I_d = (\delta i_{d1}, \delta i_{d2}, \dots, \delta i_{dn})^T$$
 (6.19′)

$$\delta I_q = (\delta i_{q1}, \delta i_{q2}, \dots, \delta i_{qn})^T$$
 (6.20′)

$$Y_{ff} = diag(y_{ff1}, y_{ff2}, ..., y_{ffn})$$
 (6.42)

$$Y_{fd} = diag(y_{fd1}, y_{fd2}, \dots, y_{fdn})$$
 (6.43)

$$Y_{df} = diag(y_{df1}, y_{df2}, \dots, y_{dfn})$$
 (6.44)

$$Y_{dd} = diag(y_{dd1}, y_{dd2}, \dots, y_{ddn})$$
 (6.45)

$$Y_{qq} = diag(y_{qq1}, y_{qq2}, ..., y_{qqn})$$
 (6.46)

$$\delta \omega_{o}^{\Psi}_{f} = (\delta \omega_{o}^{\Psi}_{f1}, \delta \omega_{o}^{\Psi}_{f2}, \dots, \delta \omega_{o}^{\Psi}_{fn})^{T}$$
 (6.47)

$$\delta \omega_{o} \Psi_{d} = (\delta \omega_{o} \psi_{d1}, \delta \omega_{o} \psi_{d2}, \dots, \delta \omega_{o} \psi_{dn})^{T}$$
 (6.21′)

$$\delta \omega_{o} \Psi_{q} = (\delta \omega_{o} \psi_{q1}, \delta \omega_{o} \psi_{q2}, \dots, \delta \omega_{o} \psi_{qn})^{T}$$
 (6.22)

Substituting (6.39) and (6.40) into (6.18) yields

$$p\delta\Omega = A_{22}\delta\Omega + A_{23}\delta\omega_{o}\Psi_{f} + A_{24}\delta\omega_{o}\Psi_{d} +$$

$$+ A_{25} \delta \omega_{o}^{\Psi} + A_{27} \delta Z_{p} + B_{22} \delta U_{p1}$$
 (6.48)

where

$$A_{23} = G_q \cdot Y_{df}$$
 (6.49)

$$A_{24} = G_q^{Y}_{dd} + H_q$$
 (6.50)

$$A_{25} = G_{d}Y_{qq} + H_{d}$$
 (6.51)

# 6.3. Differential Equations for Field Flux Linkage $(x_3)$ .

The differential equation for field flux linkage (3.68) was derived in Section 3.6

$$p\delta\psi_{fi} = \delta v_{fi} - r_{fi}\delta i_{fi}$$
 (3.68')

Introducing  $e_{fi} = x_{afi}v_{fi}/r_{fi}$  and multiplication with  $\omega_{o}$  on both sides of (3.68′) gives

$$= \omega_{o} r_{fi} / x_{afi} \delta e_{fi} - \omega_{o} r_{fi} \delta i_{fi}$$
 (6.52)

which can be written symbolically

$$p\delta\omega_{o}\Psi_{f} = -\omega_{o}R_{f}\delta I_{f} + A_{36}\delta E_{f}$$
 (6.53)

where

$$\delta \omega_{o} \Psi_{f} = (\delta \omega_{o} \Psi_{f1}, \delta \omega_{o} \Psi_{f2}, \dots, \delta \omega_{o} \Psi_{fn})^{T}$$
 (6.54)

$$\delta I_{f} = (\delta i_{f1}, \delta i_{f2}, \dots, \delta i_{fn})^{T}$$
(6.55)

$$\delta E_{f} = (\delta e_{f1}, \delta e_{f2}, \dots, \delta e_{fn})^{T}$$
(6.56)

$$R_f = diag(r_{f1}, r_{f2}, ..., r_{fn})$$
 (6.57)

$$A_{36} = \operatorname{diag}(\omega_{o} r_{f1} / x_{af1}, \omega_{o} r_{f2} / x_{af2}, \dots, \omega_{o} r_{fn} / x_{afn})$$

$$(6.58)$$

Substitution of (6.38) into (6.53) finally yields

$$P^{\delta \omega_0 \Psi_f} = A_{33}^{\delta \omega_0 \Psi_f} + A_{34}^{\delta \omega_0 \Psi_d} + A_{36}^{\delta E_f}$$
 (6.59)

where

$$A_{33} = -\omega_o R_f Y_{ff}$$
 (6.60)

$$A_{34} = -\omega_o R_f Y_{fd}$$
 (6.61)

# 6.4. Differential Equations for Armature d-Axis Flux Linkage $(x_{ij})$ .

From Section 3.6 we have

$$p^{\delta\psi}_{di} = \delta v_{di} + r_{ai}\delta i_{di} + \omega_{o}\delta\psi_{qi} + \psi_{qi}\delta\omega_{i}$$
 (3.69′)

which after multiplication with  $\boldsymbol{\omega}_{o}$  on both sides can be written symbolically

$$p\delta\omega_{o}^{\Psi}_{d} = \omega_{o}\delta V_{d} + \omega_{o}R_{a}\delta I_{d} + \omega_{o}\delta\omega_{o}^{\Psi}_{q} + \omega_{o}^{\Psi}_{q}\delta\Omega$$
 (6.62)

where

$$\delta \omega_{o} \Psi_{d} = (\delta \omega_{o} \psi_{d1}, \delta \omega_{o} \psi_{d2}, \dots, \delta \omega_{o} \psi_{dn})^{T}$$
 (6.21′)

$$\delta v_{d} = (\delta v_{d1}, \delta v_{d2}, \dots, \delta v_{dn})^{T}$$
 (6.63)

$$\delta I_{d} = (\delta i_{d1}, \delta i_{d2}, \dots, \delta i_{dn})^{T}$$
 (6.19<sup>^</sup>)

$$\delta \omega_{o}^{\Psi} = (\delta \omega_{o}^{\Psi} q_{1}, \delta \omega_{o}^{\Psi} q_{2}, \dots, \delta \omega_{o}^{\Psi} q_{n})^{T}$$
 (6.22)

$$\delta\Omega = (\delta\omega_1, \delta\omega_2, \dots, \delta\omega_n)^{\mathrm{T}}$$
 (6.16′)

$$R_a = diag(r_{a1}, r_{a2}, ..., r_{an})$$
 (6.64)

$$\Psi_{q} = \text{diag}(\psi_{q1}, \psi_{q2}, ..., \psi_{qn})$$
 (6.65)

Tis

The influence from the other generators is introduced by the  $\delta\,V_{\rm d}$  term. In Section 4.2 we derived expressions for  $\delta\,V_{\rm d}$  and  $\delta\,V_{\rm q}$ 

$$\begin{vmatrix} \delta V_{d} \\ \delta V_{q} \end{vmatrix} = \begin{vmatrix} R_{m} & -X_{m} \\ X_{m} & R_{m} \end{vmatrix} \begin{vmatrix} \delta I_{d} \\ \delta I_{q} \end{vmatrix} + \begin{vmatrix} E_{d} \\ E_{q} \end{vmatrix} \delta \Theta$$
 (4.17)

After substitution of (4.17') into (6.62) we have

$$p^{\delta \omega} \circ^{\Psi} d = \omega_{o} R_{m} \delta I_{d} - \omega_{o} X_{m} \delta I_{q} + \omega_{o} E_{d} \delta \Theta +$$

$$+ \omega_{o} R_{a} \delta I_{d} + \omega_{o} \delta \omega_{o} \Psi_{q} + \omega_{o} \Psi_{q} \delta \Omega$$

$$(6.66)$$

Combination of (6.39), (6.40) and (6.66) finally yields

$$P^{\delta \omega} \circ^{\Psi} d = A_{\mu 1}^{\delta} \circ^{\Theta} + A_{\mu 2}^{\delta \Omega} + A_{\mu 3}^{\delta \omega} \circ^{\Psi} f + A_{\mu 4}^{\delta \omega} \circ^{\Psi} d + A_{\mu 5}^{\delta \omega} \circ^{\Psi} d$$
(6.67)

where

$$A_{41} = \omega_0 E_d \tag{6.68}$$

$$A_{42} = diag(\omega_0 \psi_{q1}, \omega_0 \psi_{q2}, ..., \omega_0 \psi_{qn})$$
 (6.69)

$$A_{43} = \omega_o (R_a + R_m) Y_{df}$$
 (6.70)

$$A_{l_{1}l_{1}} = \omega_{o}(R_{a} + R_{m})Y_{dd}$$
 (6.71)

$$A_{45} = -\omega_{0}X_{m}Y_{qq} + \omega_{0}I$$
 (6.72)

# 6.5. Differential Equations for Armature q-Axis Flux Linkage (x<sub>5</sub>).

Similar algebra as in the previous section makes it possible to write

$$P^{\delta\omega}{}_{o}{}^{\psi}{}_{q} = A_{51}{}^{\delta} + A_{52}{}^{\delta\Omega} + A_{53}{}^{\delta\omega}{}_{o}{}^{\Psi}{}_{f} + A_{54}{}^{\delta\omega}{}_{o}{}^{\Psi}{}_{d} + A_{55}{}^{\delta\omega}{}_{o}{}^{\Psi}{}_{q}$$
 (6.73)

where

$$A_{51} = \omega_o E_q \tag{6.74}$$

$$A_{52} = \operatorname{diag}(-\omega_{0}\psi_{d1}, -\omega_{0}\psi_{d2}, \dots, -\omega_{0}\psi_{dn}) \qquad (6.75)$$

$$A_{53} = \omega_o X_m Y_{df}$$
 (6.76)

$$A_{54} = \omega_o X_m Y_{dd} - \omega_o I \qquad (6.77)$$

$$A_{55} = \omega_o(R_a + R_m)Y_{qq}$$
 (6.78)

# 6.6. Differential Equations for Excitation Voltages $(x_6)$ .

In Section 3.7 it was stated that the exciter system could be described by

$$p\delta v_{fi} = (-\delta v_{fi} + \delta v_{ei})/T_{ei}$$
 (3.74′)

After multiplication with  $x_{afi}/r_{fi}$  on both sides of (3.74') we have

$$p\delta e_{fi} = (-\delta e_{fi} + \delta u_{ei})/T_{ei}$$
 (6.79)

where

$$u_{ei} = x_{afi}/r_{fi}$$
 (6.80)

Equation (6.79) can be written symbolically

$$p\delta E_{f} = A_{66}\delta E_{f} + B_{61}\delta U_{e}$$
 (6.81)

where

$$E_f = (e_{f1}, e_{f2}, ..., e_{fn})^T$$
 (6.56')

$$U_e = (u_{e1}, u_{e2}, ..., u_{en})^T$$
 (6.82)

$$A_{66} = diag(-1/T_{e1}, -1/T_{e2}, ..., -1/T_{en})$$
 (6.83)

$$B_{61} = diag(1/T_{e1}, 1/T_{e2}, ..., 1/T_{en})$$
 (6.84)

# 6.7. Differential Equations for Prime Mover State Variables $(x_7)$ .

The prime movers were treated in Section 5 where the models

$$p\delta z_{hi} = a_{hi}\delta z_{hi} + b_{hi}\delta u_{hi}$$
 (5.16')

$$p\delta z_{bi} = a_{bi}\delta z_{bi} + b_{b1i}\delta u_{b1i} + b_{b2i}\delta u_{b2i}$$
 (5.25')

were obtained for the hydro plants and steam plants respectively. The models may be unified to

$$p^{\delta Z}_{p} = A_{77}^{\delta Z}_{p} + B_{72}^{\delta U}_{t} + B_{73}^{\delta U}_{f}$$
 (6.85)

where

$$\delta Z_p = (\delta z_{p1}, \delta z_{p2}, \dots, \delta z_{pn})^T$$
(6.23)

$$\delta U_{t} = (\delta u_{t1}, \delta u_{t2}, ..., \delta u_{tn})^{T}$$
 (6.24)

$$\delta U_{f} = (\delta u_{f1}, \delta u_{f2}, ..., \delta u_{fn})^{T}$$
 (6.86)

$$A_{77} = diag(a_1, a_2, ..., a_n)$$
 (6.87)

$$B_{72} = diag(b_{t1}, b_{t2}, ..., b_{tn})$$
 (6.88)

$$B_{73} = diag(b_{f1}, b_{f2}, ..., b_{fn})$$
 (6.89)

#### 6.8. Structure of the Complete System.

Equations (6.14), (6.48), (6.59), (6.67), (6.73), (6.81) and (6.85) are collected to yield

$$\begin{bmatrix} \dot{\mathbf{x}}_1 \\ \dot{\mathbf{x}}_2 \\ \dot{\mathbf{x}}_3 \\ \dot{\mathbf{x}}_4 \end{bmatrix} = \begin{bmatrix} 0 & A_{12} & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & A_{22} & A_{23} & A_{24} & A_{25} & 0 & 0 \\ 0 & 0 & A_{33} & A_{34} & A_{35} & A_{36} & 0 \\ A_{41} & A_{42} & A_{43} & A_{44} & A_{45} & 0 & 0 \\ A_{51} & A_{52} & A_{53} & A_{54} & A_{55} & 0 & 0 \\ \dot{\mathbf{x}}_7 \end{bmatrix} + \mathbf{x}_5$$

$$\begin{bmatrix} \dot{\mathbf{x}}_1 \\ \mathbf{x}_2 \\ \mathbf{x}_3 \\ \mathbf{x}_4 \end{bmatrix} + \mathbf{x}_5$$

$$\begin{bmatrix} \dot{\mathbf{x}}_1 \\ \mathbf{x}_2 \\ \mathbf{x}_3 \\ \mathbf{x}_4 \end{bmatrix} + \mathbf{x}_5$$

Thus we have a state model with 7n state variables, where n is the number of generators. In practice the time derivatives of  $\psi_d$  and  $\psi_q$  are often very small. In this  $\psi_d$  and  $\psi_q$  may be eliminated from (6.90) to yield a model with 5n state variables.

To store the entire A and B matrices we need  $70n^2$  locations. The storage requirement can be reduced if we store only the nonzero submatrices. Since  $A_{41}$ ,  $A_{43}$ ,  $A_{44}$ ,  $A_{45}$ ,  $A_{51}$ ,  $A_{53}$ ,  $A_{54}$  and  $A_{55}$  are the only submatrices with elements outside the main diagonal the required storage can be reduced to  $8n^2 + 17n$ .

The most severe numerical procedure is the inversion of  $\tilde{Y}$  to obtain  $\tilde{Z}$ . The rest of the modelling process only requires the inversion of n 3 by 3 matrices and matrix multiplication.

#### 7. REFERENCES.

- [1] Colombo, A., Redaelli, F., Ruckstuhl, G., and Vian, A.: "Determination of the Dynamic Response of Electrical Systems by Means of a Digital Program", IEEE Trans.

  Power Apparatus and Systems, Vol. PAS-87, pp. 1411-1418, June, 1968.
- [2] Ewart, D.N., and deMello, F.P.: "A Digital Computer Program for the Automatic Determination of Dynamic Stability Limits", IEEE Trans. Power Apparatus and Systems, Vol. PAS-86, pp. 867-875, July, 1967.
- [3] Gustavsson, H., Johansson, T., Sundström, L., and Ölwegård, Å.: "A Computer Program for Stability Calculations in AC Networks
  Including Voltage and Prime Mover Regulation and Containing DC Links", Proc. Power Systems Computation Conf. (Stockholm, Sweden, June-July, 1966), Rept. 5.10.
- [4] Johansson, A.: "En metod att beräkna statisk stabilitet samt några tillämpningar",

  Inst. f. El.anläggningsteknik och Elektromaskinlära, Medd. 29, June, 1967.
- [5] van Ness, J.E.: "Sensitivities of Large Multiple-Loop Control Systems", IEEE Trans. Automatic Control, Vol. AC-10, pp. 308--315, June, 1965.
- [6] Park, R.H.: "Two-Reaction Theory of Synchronous Machines, Generalized Method of Analysis Part I", AIEE Trans., Vol. 48, pp. 716-730, July, 1929.

- [7] Park, R.H.: "Two-Reaction Theory of Synchronous Machine II", AIEE Trans., Vol. 52, pp. 352-355, June, 1933.
- [8] Prabhashankar, K., and Janischewsyj, W.: "Digital Simulation of Multimachine Power Systems for Stability Studies", IEEE Trans. Power Apparatus and Systems, Vol. PAS-87, pp. 73-80, Jan., 1968.
- [9] Stanton, K.N.: "Simulating the Dynamic Behaviour of Power Systems During Large Disturbances", The Institution of Engineers, Australia, Electrical Engineering Transactions, pp. 193-199, March, 1969.
- [10] Taylor, D.G.: "Analysis of Synchronous Machines Connected to Power System Networks", Proc. IEE (London), Vol. 109, pt. C, pp. 606-610, July, 1962.
- [11] Undrill, J.M.: "Dynamic Stability Calculations for an Arbitrary Number of Interconnected Synchronous Machines", IEEE Trans.

  Power Apparatus and Systems, Vol. PAS-87, pp. 835-843, March, 1968.
- [12] von Zweygbergk, S.: "Synkronmaskinens teori för transient tillstånd", Komp., Inst. f. Elektromaskinlära, CTH, 1968.
- [13] Åström, K.J.: "Reglerteori", Almqvist & Wiksell, Stockholm, 1968.
- [14] Aström, K.J., and Eklund, K.: "A Simplified Nonlinear Model of a Drum Boiler-Turbine Unit", Report 7104, Division of Automatic Control, LTH, April, 1971.
- [15] Ölwegård, A.: Private communication.

#### APPENDIX: LIST OF SYMBOLS

### 1. General Control Theory

x state vector

u control vector

A,B system matrices

V lossfunction

 $\mathbf{Q}_1,\mathbf{Q}_2$  weighting matrices

L feedback matrix

S solution to the Riccati equation

### 2. General Power System Theory

 $P_{g}, P_{gi}$  active power generation

 $Q_{g}, Q_{gi}$  reactive power generation

 $P_d, P_{di}$  active power demand

 $Q_d, Q_{di}$  reactive power demand

P,P<sub>i</sub> active power injection

Q,Q; reactive power generation

P<sub>m</sub>,P<sub>mi</sub> mechanical power

 $\mathbf{E}_{\mathbf{f}}, \mathbf{E}_{\mathbf{f}i}$  excitation voltage or open circuit voltage

u<sub>e</sub>,u<sub>ei</sub> exciter input

u<sub>p</sub>,u<sub>pi</sub> prime mover input(s)

Ĭ	node current (complex)
V	node voltage (complex)
Y Y	node admittance matrix (complex)
${\mathtt I}_{\mathrm D}$	real part of I
$I_Q$	imaginary part of I
$v_{\rm D}$	real part of V
$v_Q$	imaginary part of V
$G_{N}$	real part of Y
$B_{N}$	imaginary part of Y

#### 3. Synchronous Machine

 $\Psi_{\rm p}, \Psi_{\rm s}, \Psi_{\rm t}, \Psi_{\rm f}$ 

flux linkages, 3-phase machine

 $i_r, i_s, i_t, i_f$ 

currents, 3-phse machine

 $v_{r}, v_{s}, v_{t}, v_{f}$ 

voltages, 3-phse machine

l xv inductances, 3-phse machine

<sup>y</sup>ösţf

 $(\psi_{r},\psi_{s},\psi_{t},\psi_{f})^{T}$ 

1 rstf  $(i_r, i_s, i_t, i_f)^T$ 

Vrstf

 $(v_r, v_s, v_t, v_f)^T$ 

Lrstf

(Lrstf)xy=1xy

ra

armature resistance, 3-phase machine

 $r_{\mathbf{f}}$ 

field resistance, 3-phse machine

Rrstf

 $diag(r_a,r_a,r_a,r_f)$ 

 $N_3$ 

number of turns/phase, 3-phase machine

И

number of turns/phase, 2-phase machine

 $\psi_{\alpha}, \psi_{\beta}, \psi_{O}, \psi_{f}$ 

flux linkages, 2-phase machine

 $i_{\alpha}, i_{\beta}, i_{o}, i_{f}$ 

currents, 2-phase machine

 $v_{\alpha}, v_{\beta}, v_{o}, v_{f}$ 

voltages, 2-phase machine

Ψαβοξέ

 $(\psi_{\alpha}, \psi_{\beta}, \psi_{o}, \psi_{f})^{\mathrm{T}}$ 

Iαβοf

 $(i_{\alpha}, i_{\beta}, i_{\circ}, i_{f})^{T}$ 

Vαβof

 $(v_{\alpha}, v_{\beta}, v_{o}, v_{f})^{T}$ 

 $c_1$ 

transformation matrix for flux linkages, currents and voltages from 3-phase to 2-phse machine

P rstf	active pover, 3-phase machine
$P_{\alpha\beta\circ f}$	active power, 2-phase machine
Lαβof	inductance matrix, 2-phse machine
$R_{\alpha \beta of}$	resistance matrix, 2-phase machine
N	number of turns/phase, dq-machine
$^{\psi_{ m d},\psi_{ m q},\psi_{ m o},\psi_{ m f}}$	flux linkages, dq-machine
$i_d, i_q, i_o, i_f$	currents, dq-machine
v <sub>d</sub> ,v <sub>q</sub> ,v <sub>o</sub> ,v <sub>f</sub>	voltages, dq-machine
Ψdqof	$(\psi_{\mathrm{d}},\psi_{\mathrm{q}},\psi_{\mathrm{o}},\psi_{\mathrm{f}})^{\mathrm{T}}$
I <sub>dq</sub> %f	$(i_d, i_q, i_o, i_f)^T$
V <sub>dqof</sub>	$(v_d, v_q, v_o, v_f)^T$
Pdqof	active power, dq-machine
C <sub>2</sub>	transformation matrix for flux linkages, currents and voltages from 2-phase to dq-machine
$^{ m L}$ dq $_{_{\circlearrowleft}}$ f	inductance matrix, dq-machine
<sup>R</sup> dqof	resistance matrix, dq-machine
E	energy
E <sub>e</sub>	energy, stored in magnetic circuits
E <sub>m</sub>	energy, stored in rotating masses
$P_{m}$	mechanical power

Pn electrical power from network  $P_{lr}$ power losses in resistances  $P_{lm}$ power losses, mechanical P<sub>ld</sub> power losses, in damper windings J moment of inertia angular velocity ω  $D_{7}$ damping coefficient, damper windings  $D_2$ damping coefficient, mechanical damping  $^{\rm M}_{\rm e}$ air-gap torque  $M_{\rm d}$ damping torque self-inductance of field winding  $L_{\mathsf{f}}$  $L_{af}$ mutual inductance between stator d-axis winding and field winding self-inductance of stator d-axis winding  $L_{d}$ self-inductance of stator q-axis winding  $(\psi_f, \psi_d, \psi_q)^T$  $(i_f, i_d, i_q)^T$ Ι see (3.61) L W see (3.43)  $T_{e}$ exciter time-constant input signal to exciter  $u_e$ 

#### 4. Transmission Network

 $\mathbf{v}_{\mathrm{D}}, \mathbf{v}_{\mathrm{Di}}$ 

network terminal voltages expressed with reference to network reference axes

 $v_Q, v_{Qi}$ 

 $\mathtt{V}_{\mathrm{D}}$ 

 $(v_{\underline{\text{D1}}},v_{\underline{\text{D2}}},\dots,v_{\underline{\text{Dn}}})^{\text{T}}$ 

 $v_{Q}$ 

 $(v_{Q1}, v_{Q2}, \ldots, v_{Qn})^T$ 

 $\theta_{i}$ 

angular displacement of the (d,q) axes of machine i with respect to (D,Q) axes of the network. (Fig. 4.1)

Θ

$$(\theta_1, \theta_2, \ldots, \theta_n)^T$$

 $i_{D}, i_{Di}$ 

network terminal currents expressed with reference to network reference axes

i<sub>Q</sub>,i<sub>Qi</sub>

 $g_{ij}$ 

real component of a network self- or mutual admittance

<sup>b</sup>ij

imaginary component of a network self- or mutual admittance

 $R_{M}$ 

see (4.15)

 $X_{M}$ 

see (4.16)

 $E_{D}$ 

see (4,18)

 $E_Q$ 

see (4.19)

### 5. Hydro Turbines

q flow of water

p density of water

vout velocity of water at outlet

a outlet area

 $\eta(\omega)$  speed dependent efficiency

v<sub>t</sub> velocity of water in the dash-tube

L length of the dashtube

g constant of gravity

h water head

A dash-tube area

 $v_{\mathsf{tmax}}$  maximum velocity of water in the dash-tube

state variable (v<sub>t</sub>/v<sub>tmax</sub>)

t control variable (a/a max)

a maximum outlet area

time-constant, water-system

### 6. Boilers and Steam Turbines

₽

steam preassure

u-

steam valve setting

 $\mathbf{u}_{2}$ 

fuel flow

 $u_3$ 

feedwater flow

 $\alpha_1, \alpha_2, \alpha_3, \alpha_{\underset{i_i}{\mu}}, \alpha_5$ 

parameters, steam plant

q

steam flow

z p state variable (p/p<sub>max</sub>)

 $u_{+}$ 

normalized steam valve setting

 $\mathbf{u}_{\mathbf{f}}$ 

normalized fuel flow

### 7. Operators

p(.)

differential operator p(x)=dx/dt

δ(.)

incremental operator  $\delta(x)=x-x_0$ 

 $(.)^{T}$ 

transposition of a matrix

diag(
$$\lambda_1, \lambda_2, \ldots, \lambda_n$$
)

$$\Lambda$$
=diag( $\lambda_1, \lambda_2, \ldots, \lambda_n$ )